Physics-Based Deep Learning for Fiber-Optic Communication System

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Thank You!



Henry D. Pfister Duke



Christoffer Fougstedt Chalmers (now: Ericsson)



Lars Svensson Chalmers



Per Larsson-Edefors Chalmers



Rick M. Bütler TU/e (now: TU Delft)



Gabriele Liga TU/e



Alex Alvarado TU/e



Sjoerd van der Heide TU/e



Vinícius Oliari TU/e



Sebastiaan Goossens TU/e

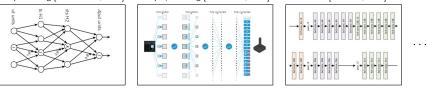


Menno van den Hout TU/e



Chigo Okonkwo TU/e



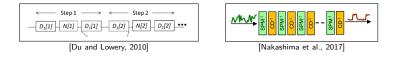


Multi-layer neural networks: impressive performance, countless applications





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Split-step methods for solving the propagation equation in fiber-optics

Machine Learning	Physics-Based Models	Learned DBP	Conclusions	CHALMERS
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1. show that multi-layer neural networks and the split-step method have the same functional form: both alternate linear and pointwise nonlinear steps

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- propose a physics-based machine-learning approach based on parameterizing the split-step method (no black-box neural networks)
- 3. revisit hardware-efficient nonlinear equalization via learned digital backpropagation

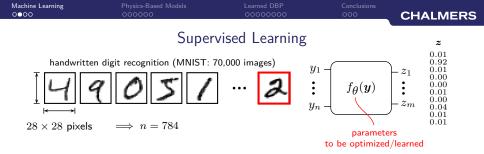


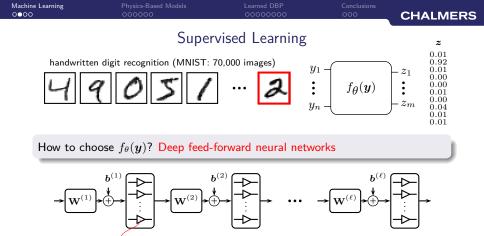
- 1. Machine Learning and Neural Networks for Communications
- 2. Physics-Based Machine Learning for Fiber-Optic Communications
- 3. Learned Digital Backpropagation
- 4. Conclusions



1. Machine Learning and Neural Networks for Communications

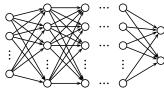
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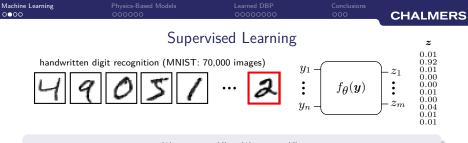


activation function

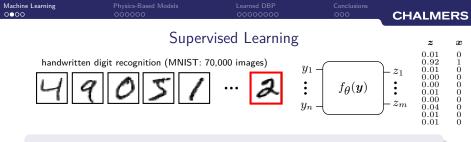




equivalent graph representation



How to optimize $\theta = \{ W^{(1)}, ..., W^{(\ell)}, b^{(1)}, ..., b^{(\ell)} \}$?



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Given a data set $\mathcal{D} = \{(y^{(i)}, x^{(i)})\}_{i=1}^N$, where $y^{(i)}$ are model inputs and $x^{(i)}$ are labels, we iteratively minimize

$$\frac{1}{|\mathcal{B}_k|} \sum_{(\boldsymbol{y}, \boldsymbol{x}) \in \mathcal{B}_k} \mathcal{L}(f_{\theta}(\boldsymbol{y}), \boldsymbol{x}) \triangleq g(\theta) \qquad \qquad \text{using} \quad \begin{array}{l} \theta_{k+1} = \theta_k - \lambda \nabla_{\theta} g(\theta_k) \\ \text{stochastic gradient descent} \end{array}$$

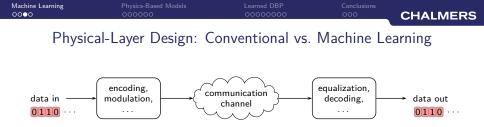
- $\mathcal{B}_k \subset \mathcal{D}$ and $|\mathcal{B}_k|$ is called the batch (or minibatch) size
- Typical loss function: mean squared error $\mathcal{L}(a,b) = \|a b\|^2$ (regression)
- λ is called the step size or learning rate

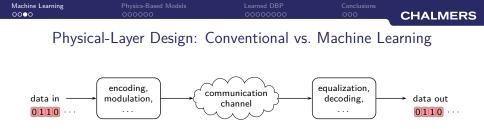
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channel

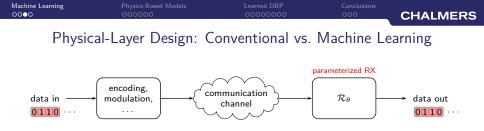


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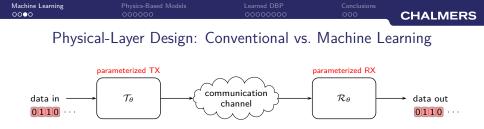
- Model deficiency: no good model might be available
- Algorithm deficiency: infeasible algorithms may require simplifications



Conventional: handcrafted DSP blocks based on mathematical modeling

- Model deficiency: no good model might be available
- Algorithm deficiency: infeasible algorithms may require simplifications
- Use function approximators and learn parameter configurations θ from data

[Shen and Lau, 2011], Fiber nonlinearity compensation using extreme learning machine for DSP-based ..., (*OECC*) [Giacoumidis et al., 2015], Fiber nonlinearity-induced penalty reduction in CO-OFDM by ANN-based ..., (*Opt. Lett.*)



- Model deficiency: no good model might be available
- Algorithm deficiency: infeasible algorithms may require simplifications
- Use function approximators and learn parameter configurations θ from data
- Joint transmitter-receiver learning via autoencoder [O'Shea and Hoydis, 2017]

[[]Karanov et al., 2018], End-to-end deep learning of optical fiber communications (J. Lightw. Technol.)

[[]Li et al., 2018], Achievable information rates for nonlinear fiber communication via end-to-end autoencoder learning, (ECOC)



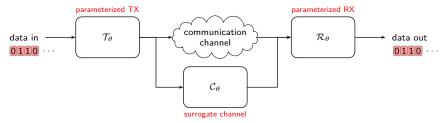
surrogate channel
 Conventional: handcrafted DSP blocks based on mathematical modeling

 \mathcal{C}_{θ}

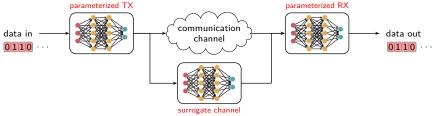
- Model deficiency: no good model might be available
- Algorithm deficiency: infeasible algorithms may require simplifications
- Use function approximators and learn parameter configurations θ from data
- Joint transmitter-receiver learning via autoencoder [O'Shea and Hoydis, 2017]
- Surrogate channel models for gradient-based TX training

[[]O'Shea et al., 2018], Approximating the void: Learning stochastic channel models from observation with variational GANs, (arXiv) Ye et al., 2018], Channel agnostic end-to-end learning based communication systems with conditional GAN, (arXiv)





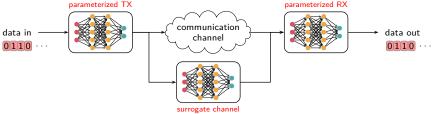




Using (deep) neural networks for $\mathcal{T}_{\theta}, \mathcal{R}_{\theta}, \mathcal{C}_{\theta}$? Possible, but . . .

- How to choose the network architecture (#layers, activation function)?
- How to limit the number of parameters (complexity)?
- How to interpret the solutions? Any insight gained?
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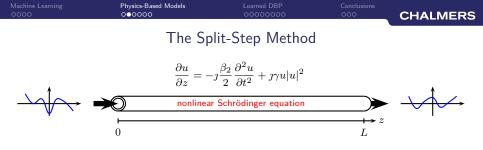
Our contribution: designing "neural-network-like" machine-learning models by exploiting the underlying physics of the propagation.

Machine Learning	Physics-Based Models	Learned DBP	Conclusions	CHALMERS
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Outline				

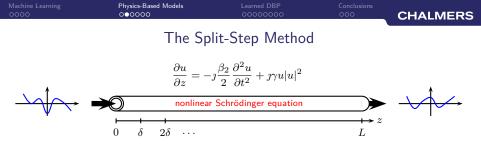
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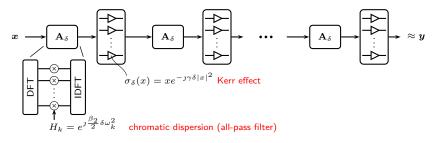
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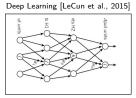
• Deterministic channel model: partial differential equation



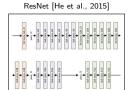
- Deterministic channel model: partial differential equation
- Split-step method with M steps ($\delta = L/M$):



Machine Learning 0000	Physics-Based Models 00●000	Learned DBP 00000000	Conclusions 000	CHALMERS





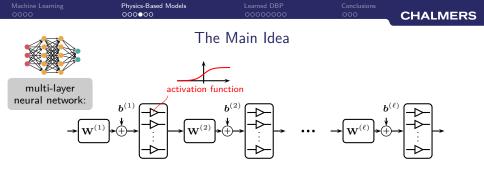


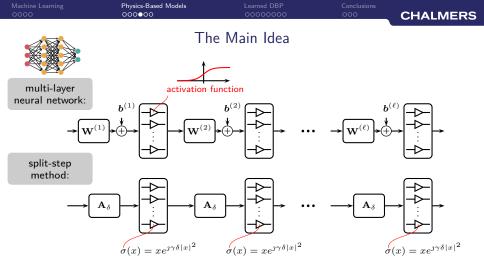




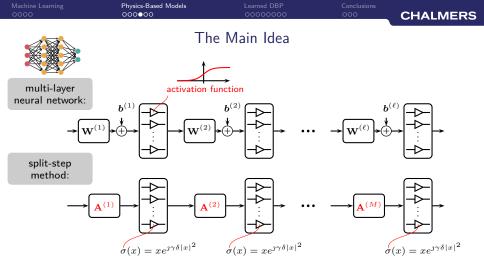


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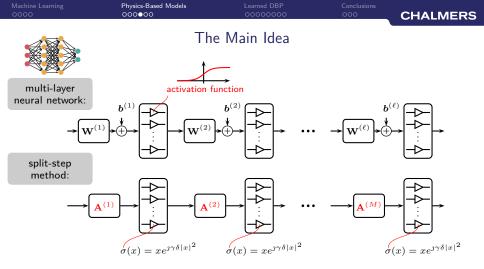
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[Häger & Pfister, 2018], Nonlinear Interference Mitigation via Deep Neural Networks, (OFC)

[Häger & Pfister, 2021], Physics-Based Deep Learning for Fiber-Optic Communication Systems, IEEE J. Sel. Areas Commun.

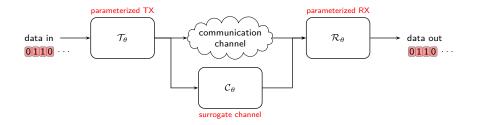


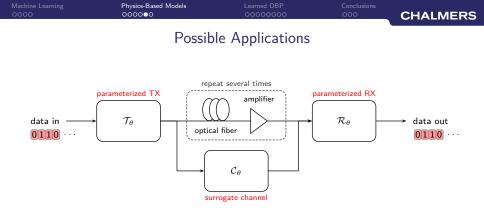
- This almost looks like a deep neural net!
- Parameterize all linear steps: f_{θ} with $\theta = {\mathbf{A}^{(1)}, \dots, \mathbf{A}^{(M)}}$
- Special cases: step-size optimization, nonlinear operator "placement",

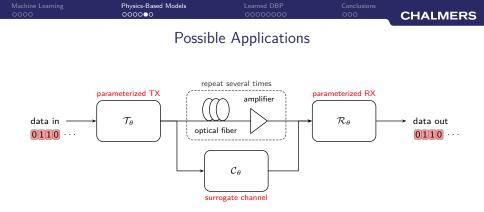
[Häger & Pfister, 2018], Nonlinear Interference Mitigation via Deep Neural Networks, (OFC) [Häger & Pfister, 2021], Physics-Based Deep Learning for Fiber-Optic Communication Systems, IEEE J. Sel. Areas Commun.



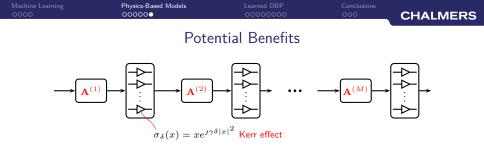
Possible Applications







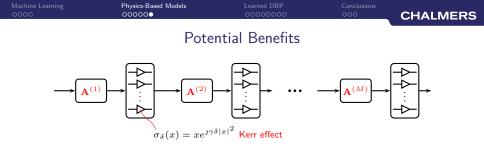
- Channel C_θ: fine-tune model based on experimental data, reduce simulation time [Leibrich and Rosenkranz, 2003], [Li et al., 2005]
- Receiver \mathcal{R}_{θ} : nonlinear equalization (focus in this talk)
- Transmitter T_θ: digital pre-distortion [Essiambre and Winzer, 2005], [Roberts et al., 2006], "split" nonlinearity compensation [Lavery et al., 2016]



• How to choose the network architecture (#layers, activation function)?

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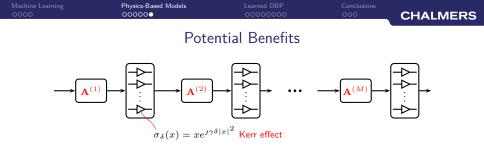
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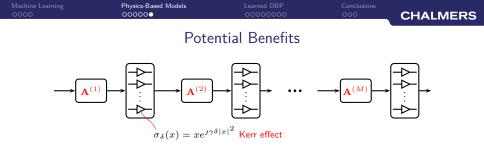
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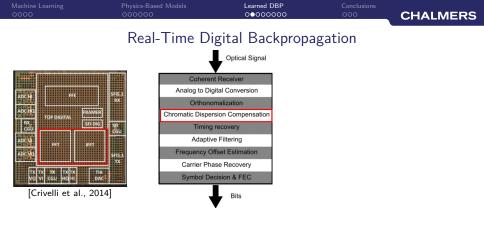


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 - Learned parameter configurations are interpretable
 - Satisfactory explanations for benefits over previous handcrafted solutions



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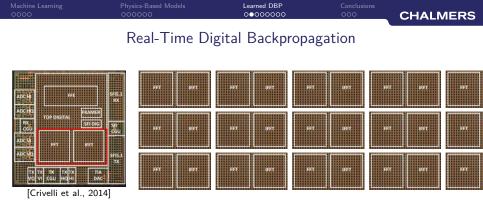
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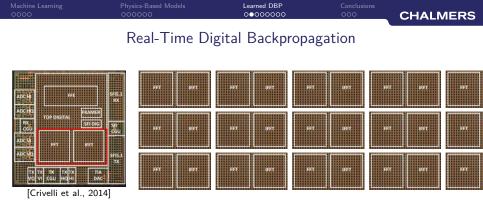
[Crivelli et al., 2014]

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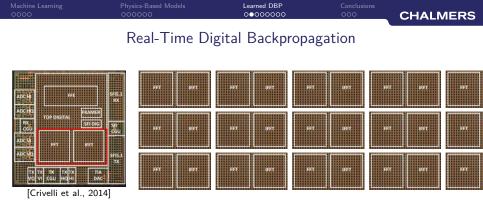
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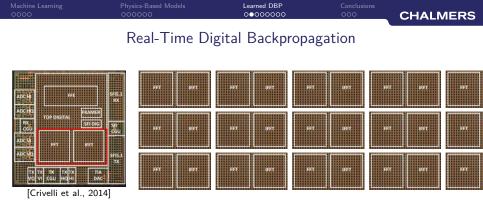
- Invert a PDE in real time [Essiambre and Winzer, 2005], [Roberts et al., 2006], [Li et al., 2008], [Ip and Kahn, 2008]: widely considered to be impractical
- Complexity increases with the number of steps M ⇒ reduce M as much as possible (see, e.g., [Du and Lowery, 2010], [Rafique et al., 2011], [Li et al., 2011], [Yan et al., 2011], [Napoli et al., 2014], [Secondini et al., 2016], ...)



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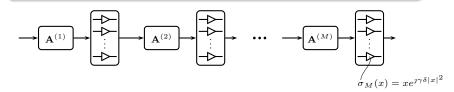
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Our approach: many steps but model compression

Joint optimization, pruning, and quantization of all linear steps

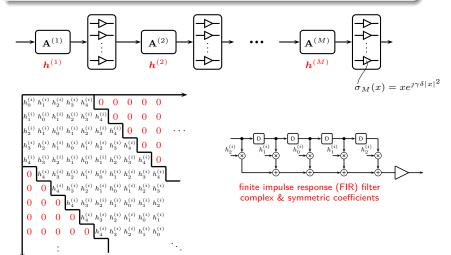


TensorFlow implementation of the computation graph $f_{\theta}(\boldsymbol{y})$:



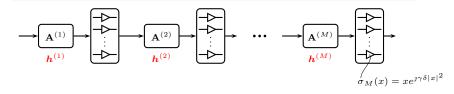


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Deep learning of all FIR filter coefficients $\theta = {\mathbf{h}^{(1)}, \dots, \mathbf{h}^{(M)}}$:

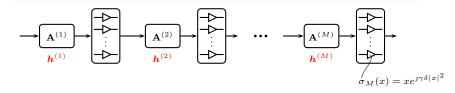
$$\min_{\theta} \sum_{i=1}^{N} \mathsf{Loss}(f_{\theta}(\boldsymbol{y}^{(i)}), \boldsymbol{x}^{(i)}) \triangleq g(\theta)$$
mean squared error

using $\theta_{k+1} = \theta_k - \lambda \nabla_{\theta} g(\theta_k)$

Adam optimizer, fixed learning rate



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Mean squared error Adam optimizer, fixed learning rate

Iteratively prune (set to 0) outermost filter taps during gradient descent

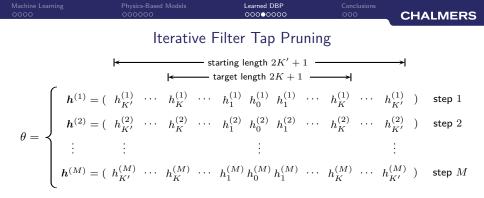
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Iterative Filter Tap Pruning

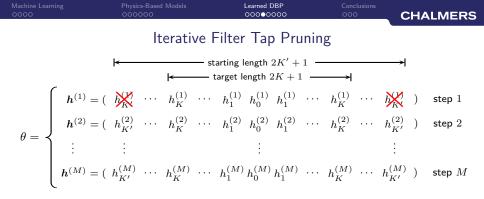
$$heta = \left\{egin{array}{c} oldsymbol{h}^{(1)} & & \ oldsymbol{h}^{(2)} & & \ dots & & \ dots & & \ oldsymbol{h}^{(M)} & & \ oldsymbol{h}^{(M)$$

$$\theta = \begin{cases} h^{(1)} = (\ h^{(1)}_{K'} \ \cdots \ h^{(1)}_{K} \ \cdots \ h^{(1)}_{K} \ h^{(1)}_{K'} \ h^{(1)}_$$

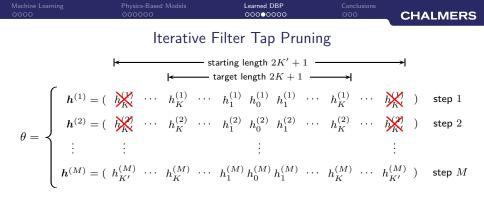
• Initially: constrained least-squares coefficients (LS-CO) [Sheikh et al., 2016]



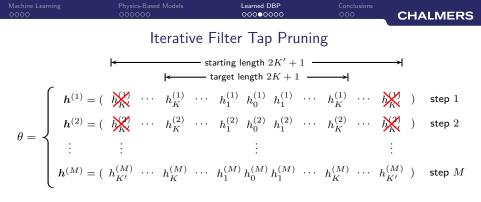
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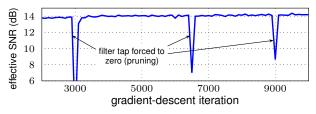
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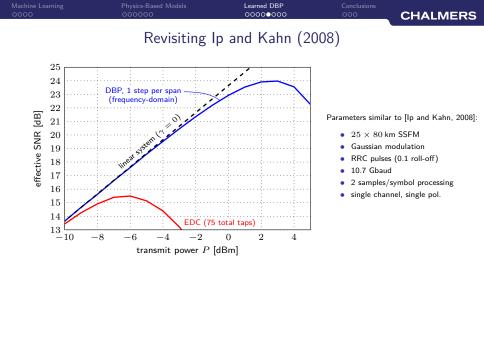


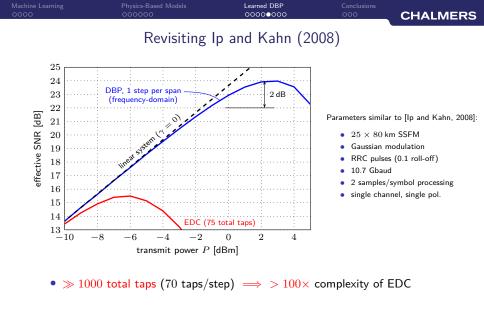
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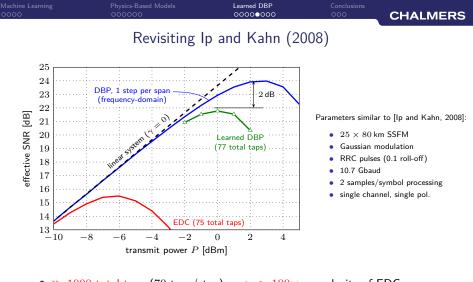


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- Typical learning curve:

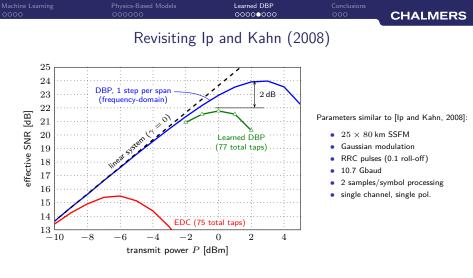








- $\gg 1000$ total taps (70 taps/step) $\implies > 100 \times$ complexity of EDC
- Learned approach uses only 77 total taps: alternate 5 and 3 taps/step and use different filter coefficients in all steps [Häger and Pfister, 2018a]

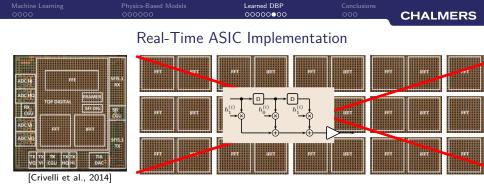


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- Can outperform "ideal DBP" in the nonlinear regime [Häger and Pfister, 2018b]





[[]Fougstedt et al., 2017]. Time-domain digital back propagation: Algorithm and finite-precision implementation aspects, (OFC) Fougstedt et al., 2018]. ASIC implementation of time-domain digital back propagation for coherent receivers, (PTL) Sherborne et al., 2018]. On the impact of fixed point hardware for optical fibre nonlinearity compensation algorithms, (JLT)



• Our linear steps are very short symmetric FIR filters (as few as 3 taps)



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- 28-nm ASIC at 416.7 MHz clock speed (40 GHz signal)
 - Only 5-6 bit filter coefficients via learned quantization
 - Hardware-friendly nonlinear steps (Taylor expansion)
 - All FIR filters are fully reconfigurable

[[]Fougstedt et al., 2018], ASIC implementation of time-domain digital backpropagation with deep-learned chromatic dispersion filters, (ECOC)

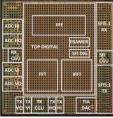
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Machine Learning Physics-Based Models Learned DBP Conclusions 0000 000000 000 000 CHALME

Real-Time ASIC Implementation



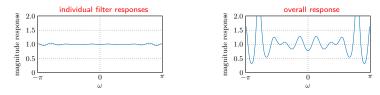
- [Crivelli et al., 2014]
 - Our linear steps are very short symmetric FIR filters (as few as 3 taps)
 - 28-nm ASIC at 416.7 MHz clock speed (40 GHz signal)
 - Only 5-6 bit filter coefficients via learned quantization
 - Hardware-friendly nonlinear steps (Taylor expansion)
 - All FIR filters are fully reconfigurable
 - $< 2 \times$ power compared to EDC [Crivelli et al., 2014, Pillai et al., 2014]

[[]Fougstedt et al., 2018], ASIC implementation of time-domain digital backpropagation with deep-learned chromatic dispersion filters, (ECOC)



Why Does The Learning Approach Work?

Previous work: design a single filter or filter pair and use it repeatedly. \implies Good overall response only possible with very long filters.



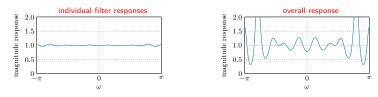
From [Ip and Kahn, 2009]:

- "We also note that [...] 70 taps, is much larger than expected"
- "This is due to amplitude ringing in the frequency domain"
- "Since backpropagation requires multiple iterations of the linear filter, amplitude distortion due to ringing accumulates (Goldfarb & Li, 2009)"



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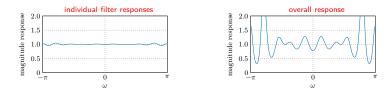
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- "This is due to amplitude ringing in the frequency domain"
- "Since backpropagation requires multiple iterations of the linear filter, amplitude distortion due to ringing accumulates (Goldfarb & Li, 2009)"

The learning approach uncovered that there is no such requirement! [Lian, Häger, Pfister, 2018]. What can machine learning teach us about communications? (*ITW*)



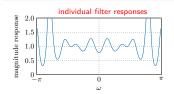
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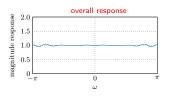
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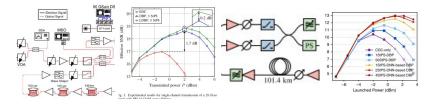
Sacrifice individual filter accuracy, but different response per step.

 \implies Good overall response even with very short filters by joint optimization.









Training with real-world data sets including presence of various hardware impairments (phase noise, timing error, frequency offset, etc.)

- [Oliari et al., 2020], Revisiting Efficient Multi-step Nonlinearity Compensation with Machine Learning: An Experimental Demonstration, (J. Lightw. Technol.)
- [Sillekens et al., 2020], Experimental Demonstration of Learned Time-domain Digital Back-propagation, (*Proc. IEEE Workshop on Signal Processing Systems*)
- [Fan et al., 2020], Advancing Theoretical Understanding and Practical Performance of Signal Processing for Nonlinear Optical Communications through Machine Learning, (Nat. Commun.)
- [Bitachon et al., 2020], Deep learning based Digital Back Propagation Demonstrating SNR gain at Low Complexity in a 1200 km Transmission Link, (*Opt. Express*)

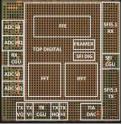


- 1. Machine Learning and Neural Networks for Communications
- 2. Physics-Based Machine Learning for Fiber-Optic Communications
- 3. Learned Digital Backpropagation
- 4. Conclusions





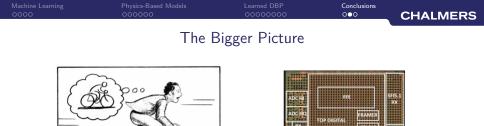
Figure 1. A World Model, from Scott McCloud's Understanding Comics. (McCloud, 1993; E, 2012)



[Crivelli et al., 2014]

• Optical receivers build models of their "environment"

[[]Ha & Schmidhuber, 2018], "World Models", arXiv:1803.10122 [cs.LG]



Interpretable physics-based "multi-layer" models for machine learning can

[Crivelli et al., 2014]

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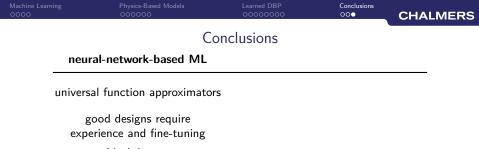
Figure 1. A World Model, from Scott McCloud's Understanding Comics. (McCloud, 1993; E, 2012)

Optical receivers build models of their "environment"
Currently these models are linear and/or rigid (non-adaptive)

be obtained by exploiting our existing domain knowledge

[[]Ha & Schmidhuber, 2018], "World Models", arXiv:1803.10122 [cs.LG]





black boxes, difficult to "open"

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[Häger & Pfister, 2020], "Physics-Based Deep Learning for Fiber-Optic Communication Systems", in *IEEE J. Sel. Areas Commun.* (to appear), see https://arxiv.org/abs/2010.14258 Code: https://github.com/chaeger/LDBP

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Conclusions						
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	good designs require experience and fine-tuning		relies on domain knowledge (algorithms, physics,)			
	black boxes, difficult to "open"	familiar building blocks (e.g., FIR filters) can enable interpretability				
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Coue	. https://github.com/chaeger/LDBr					
Thank you!						
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References I

Crivelli, D. E., Hueda, M. R., Carrer, H. S., Del Barco, M., López, R. R., Gianni, P., Finochietto, J.,

Swenson, N., Voois, P., and Agazzi, O. E. (2014).

Architecture of a single-chip 50 Gb/s DP-QPSK/BPSK transceiver with electronic dispersion compensation for coherent optical channels.

IEEE Trans. Circuits Syst. I: Reg. Papers, 61(4):1012-1025.



Du, L. B. and Lowery, A. J. (2010).

Improved single channel backpropagation for intra-channel fiber nonlinearity compensation in long-haul optical communication systems.

Opt. Express, 18(16):17075-17088.



Essiambre, R.-J. and Winzer, P. J. (2005).

Fibre nonlinearities in electronically pre-distorted transmission. In Proc. European Conf. Optical Communication (ECOC), Glasgow, UK.



Häger, C. and Pfister, H. D. (2018a).

Deep learning of the nonlinear Schrödinger equation in fiber-optic communications. In Proc. IEEE Int. Symp. Information Theory (ISIT), Vail, CO.



Häger, C. and Pfister, H. D. (2018b).

Nonlinear interference mitigation via deep neural networks. In Proc. Optical Fiber Communication Conf. (OFC), San Diego, CA.



He, K., Zhang, X., Ren, S., and Sun, J. (2015).

Deep residual learning for image recognition.

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References II



Ip, E. and Kahn, J. M. (2008).

Compensation of dispersion and nonlinear impairments using digital backpropagation. *J. Lightw. Technol.*, 26(20):3416–3425.



Ip, E. and Kahn, J. M. (2009).

Nonlinear impairment compensation using backpropagation. In Optical Fiber New Developments, Chapter 10. IntechOpen, London, UK.



Lavery, D., Ives, D., Liga, G., Alvarado, A., Savory, S. J., and Bayvel, P. (2016).

The benefit of split nonlinearity compensation for single-channel optical fiber communications. *IEEE Photon. Technol. Lett.*, 28(17):1803–1806.



LeCun, Y., Bengio, Y., and Hinton, G. (2015).

Deep learning. Nature, 521(7553):436-444.



Leibrich, J. and Rosenkranz, W. (2003).

Efficient numerical simulation of multichannel WDM transmission systems limited by XPM. *IEEE Photon. Technol. Lett.*, 15(3):395–397.



Li, L., Tao, Z., Dou, L., Yan, W., Oda, S., Tanimura, T., Hoshida, T., and Rasmussen, J. C. (2011). Implementation efficient nonlinear equalizer based on correlated digital backpropagation. In *Proc. Optical Fiber Communication Conf. (OFC)*, Los Angeles, CA.



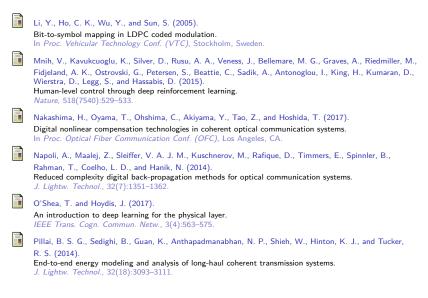
Li, X., Chen, X., Goldfarb, G., Mateo, E., Kim, I., Yaman, F., and Li, G. (2008).

Electronic post-compensation of WDM transmission impairments using coherent detection and digital signal processing.

Opt. Express, 16(2):880-888.

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References III



References IV

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Rafique, D., Zhao, J., and Ellis, A. D. (2011).

Digital back-propagation for spectrally efficient wdm 112 gbit/s pm m-ary qam transmission. *Opt. Express*, 19(6):5219–5224.



Roberts, K., Li, C., Strawczynski, L., O'Sullivan, M., and Hardcastle, I. (2006).

Electronic precompensation of optical nonlinearity. *IEEE Photon. Technol. Lett.*, 18(2):403–405.



Secondini, M., Rommel, S., Meloni, G., Fresi, F., Forestieri, E., and Poti, L. (2016). Single-step digital backpropagation for nonlinearity mitigation. *Photon. Netw. Commun.*, 31(3):493–502.



Sheikh, A., Fougstedt, C., Graell i Amat, A., Johannisson, P., Larsson-Edefors, P., and Karlsson, M. (2016). Dispersion compensation FIR filter with improved robustness to coefficient quantization errors. *J. Lightw. Technol.*, 34(22):5110–5117.



Yan, W., Tao, Z., Dou, L., Li, L., Oda, S., Tanimura, T., Hoshida, T., and Rasmussen, J. C. (2011). Low complexity digital perturbation back-propagation. In Proc. European Conf. Optical Communication (ECOC), page Tu.3.A.2, Geneva, Switzerland.