

Model-Based Machine Learning for Joint Digital Backpropagation and PMD Compensation

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Gabriele Liga⁽³⁾, Alex Alvarado⁽³⁾**

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⁽²⁾Department of Electrical and Computer Engineering, Duke University, USA

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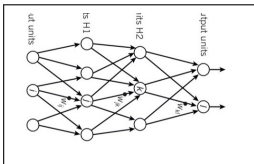
Optical Fiber Communications Conference (OFC)
San Diego, USA, March 11, 2020



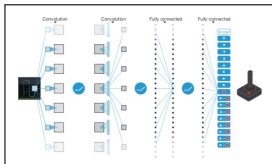
CHALMERS



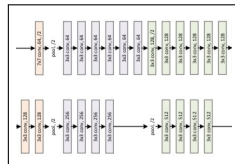
Deep Learning [LeCun et al., 2015]



Deep Q-Learning [Mnih et al., 2015]



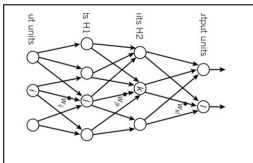
ResNet [He et al., 2015]



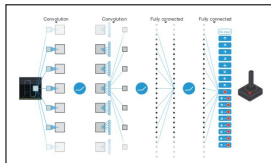
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Multi-layer neural networks: impressive performance, countless applications

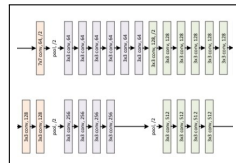
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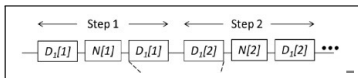


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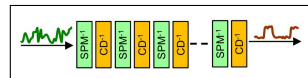


...

Multi-layer neural networks: impressive performance, countless applications



[Du and Lowery, 2010]



[Nakashima et al., 2017]

Split-step methods for solving the propagation equation in fiber-optics

Agenda

In this talk, we ...

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1. show that **multi-layer neural networks** and the **split-step method** have the same functional form: both alternate **linear** and **pointwise nonlinear** steps

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2. propose a **model-based machine-learning** approach based on **parameterizing** the split-step method (**no black-box** neural networks)

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In this talk, we ...

1. show that **multi-layer neural networks** and the **split-step method** have the same functional form: both alternate **linear** and **pointwise nonlinear** steps
2. propose a **model-based machine-learning** approach based on **parameterizing** the split-step method (**no black-box** neural networks)
3. revisit **hardware-efficient digital backpropagation** combined with **distributed** compensation of polarization mode dispersion

Outline

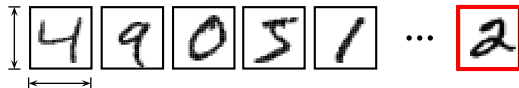
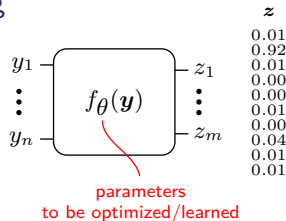
1. Machine Learning and Neural Networks for Communications
2. Model-Based Machine Learning for Fiber-Optic Communications
3. Learned Digital Backpropagation
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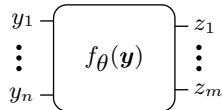
Supervised Learning

handwritten digit recognition (MNIST: 70,000 images)

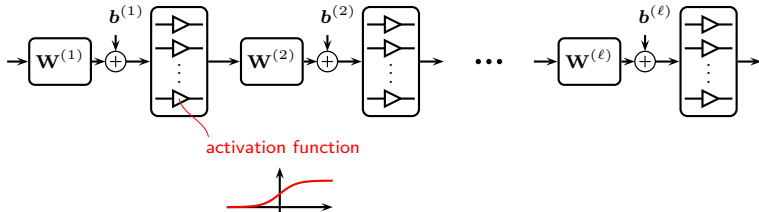
 28×28 pixels $\Rightarrow n = 784$ 

Supervised Learning

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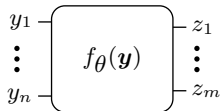

 z
 0.01
 0.92
 0.01
 0.00
 0.00
 0.01
 0.00
 0.04
 0.01
 0.01

How to choose $f_{\theta}(y)$? **Deep feed-forward neural networks**

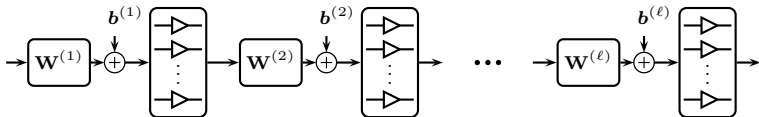


Supervised Learning

handwritten digit recognition (MNIST: 70,000 images)



| z | x |
|------|-----|
| 0.01 | 0 |
| 0.92 | 1 |
| 0.01 | 0 |
| 0.00 | 0 |
| 0.00 | 0 |
| 0.01 | 0 |
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| 0.01 | 0 |

How to choose $f_\theta(\mathbf{y})$? **Deep feed-forward neural networks**How to optimize $\theta = \{\mathbf{W}^{(1)}, \dots, \mathbf{W}^{(\ell)}, \mathbf{b}^{(1)}, \dots, \mathbf{b}^{(\ell)}\}$? **Deep learning**

$$\min_{\theta} \sum_{i=1}^N \text{Loss}(f_{\theta}(\mathbf{y}^{(i)}), \mathbf{x}^{(i)}) \triangleq g(\theta) \quad \text{using} \quad \theta_{k+1} = \theta_k - \lambda \nabla_{\theta} g(\theta_k) \quad (1)$$

mean squared error
cross-entropy, ...

stochastic gradient descent,
RMSProp, Adam, ...

Machine Learning for Physical-Layer Communications

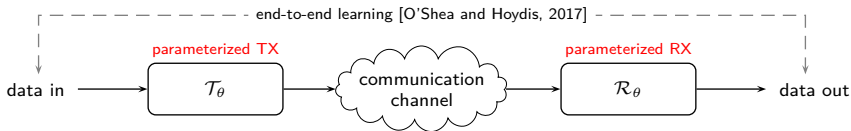


Machine Learning for Physical-Layer Communications



-
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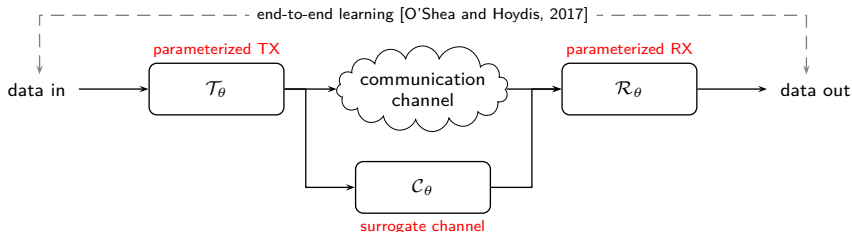
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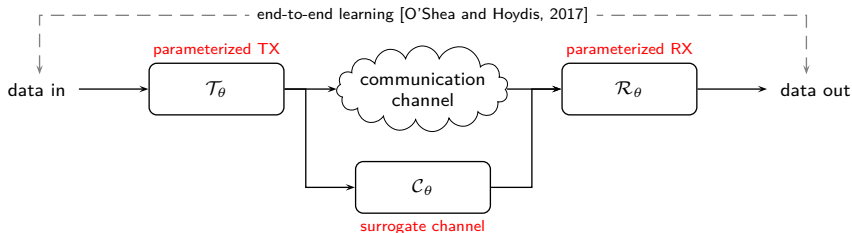


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[O'Shea et al., 2018], Approximating the void: Learning stochastic channel models from observation with variational GANs, (*arXiv*)
[Ye et al., 2018], Channel agnostic end-to-end learning based communication systems with conditional GAN, (*arXiv*)
...

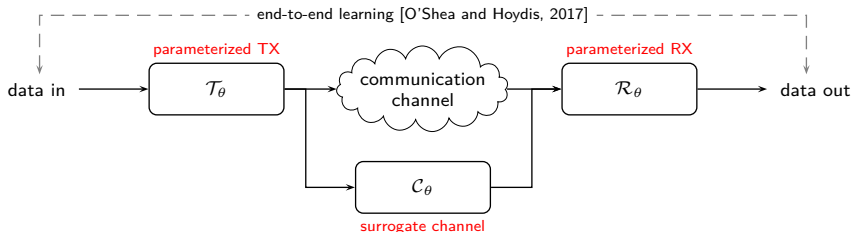
Machine Learning for Physical-Layer Communications



Using neural networks for $\mathcal{T}_\theta, \mathcal{R}_\theta, \mathcal{C}_\theta$

- How to choose **network architecture** (#layers, activation function)?
- How to **initialize** parameters?
- How to **interpret** solutions? Any **insight** gained?
- ...

Machine Learning for Physical-Layer Communications



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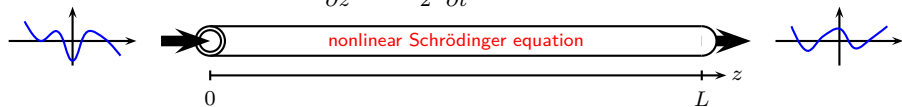
Model-based learning: sparse signal recovery [Gregor and Lecun, 2010], [Borgerding and Schniter, 2016], neural belief propagation [Nachmani et al., 2016], radio transformer networks [O'Shea and Hoydis, 2017], ...

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The Split-Step Method

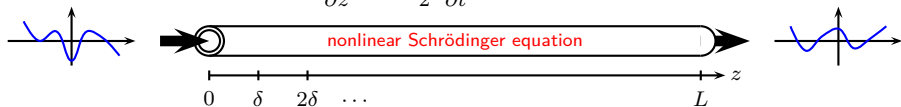
$$\frac{\partial u}{\partial z} = -j\frac{\beta_2}{2} \frac{\partial^2 u}{\partial t^2} + j\gamma u|u|^2$$



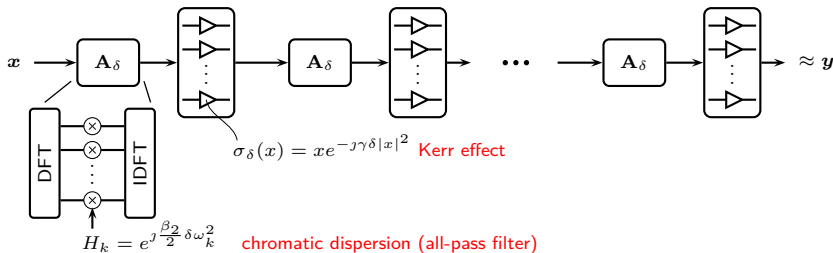
- **Deterministic channel model:** partial differential equation

The Split-Step Method

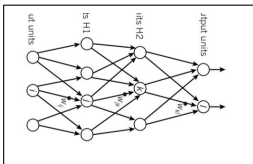
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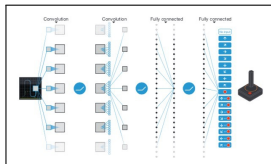
- **Deterministic channel model:** partial differential equation
- **Split-step method** with M steps ($\delta = L/M$):



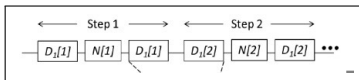
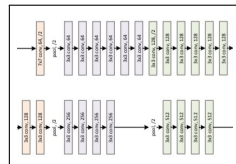
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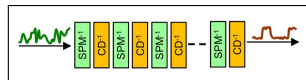
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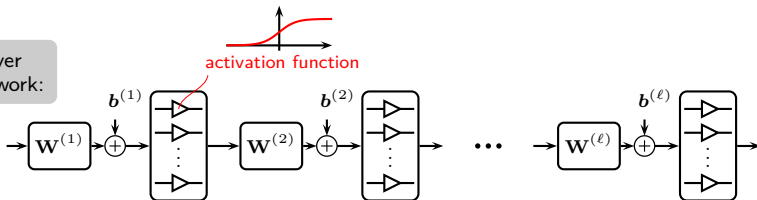
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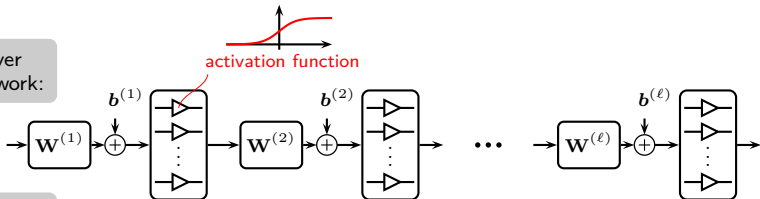
Parameterizing the Split-Step Method

multi-layer
neural network:

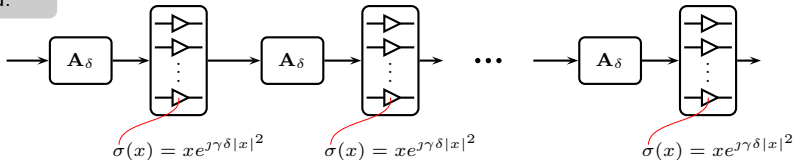


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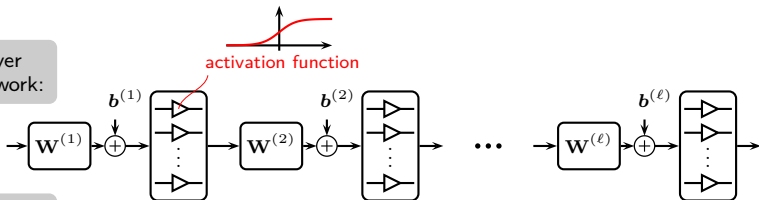


split-step
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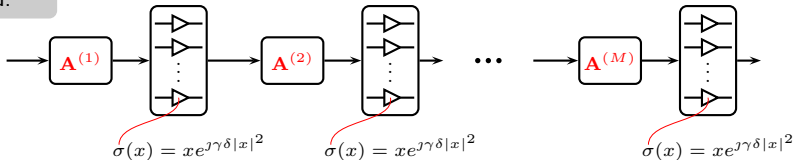


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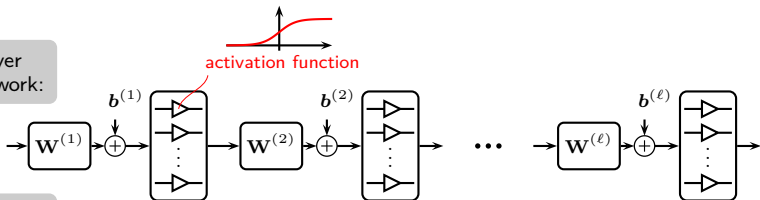


[Häger & Pfister, 2018], Nonlinear Interference Mitigation via Deep Neural Networks, (OFC)

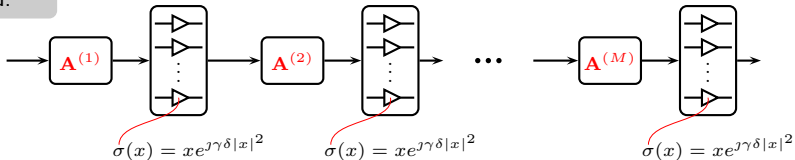
[Häger & Pfister, 2018], Deep Learning of the Nonlinear Schrödinger Equation in Fiber-Optic Communications, (ISIT)

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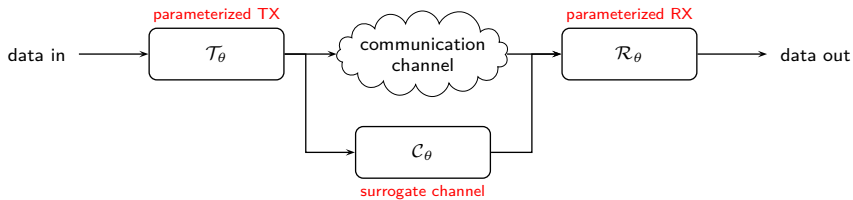


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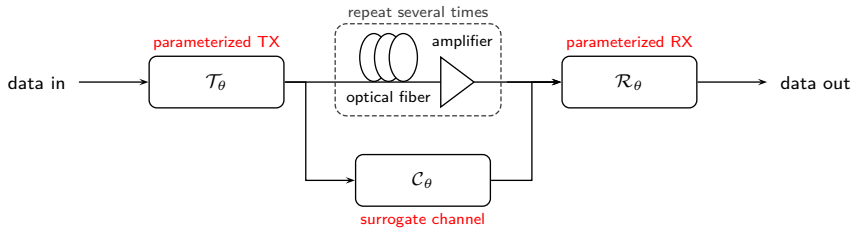


- **Parameterized model** f_θ with $\theta = \{\mathbf{A}^{(1)}, \dots, \mathbf{A}^{(M)}\}$

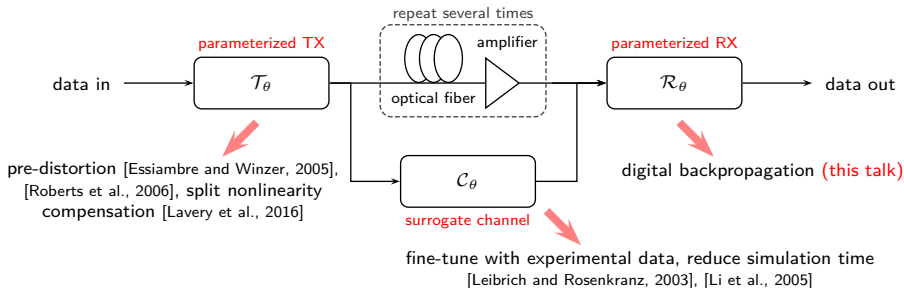
Possible Applications



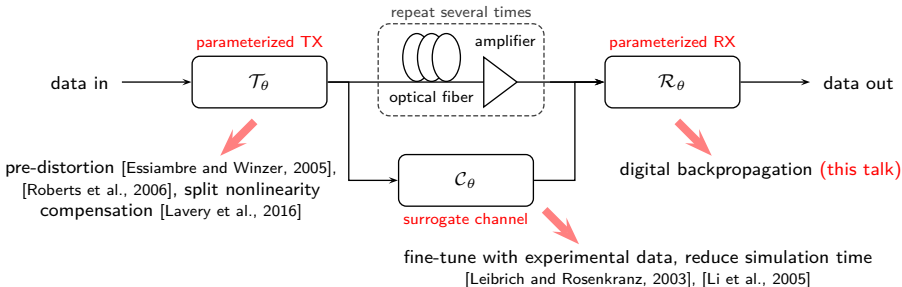
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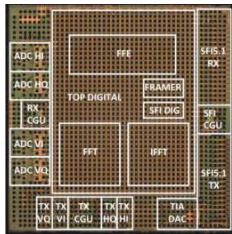
Model-based learning approaches

- How to choose **network architecture** (#layers, activation function)? ✓
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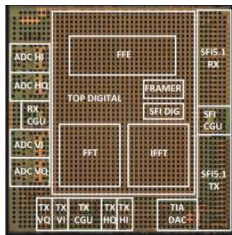
Real-Time Digital Backpropagation



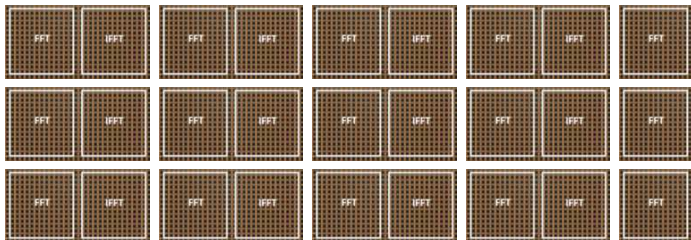
[Crivelli et al., 2014]

- Invert a partial differential equation **in real time** ([Paré et al., 1996], [Essiambre and Winzer, 2005], [Roberts et al., 2006], [Li et al., 2008], [Ip and Kahn, 2008])

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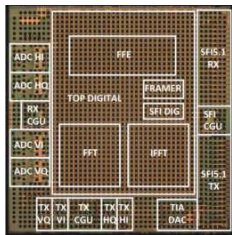


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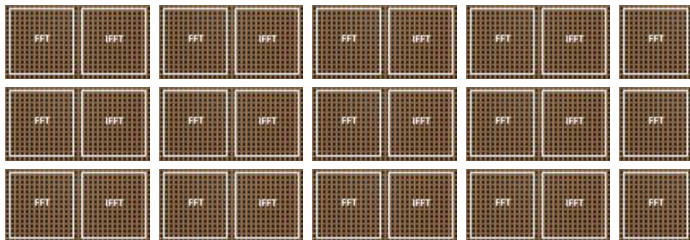


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Real-Time Digital Backpropagation



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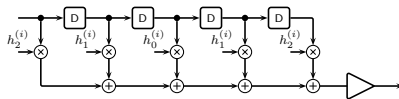
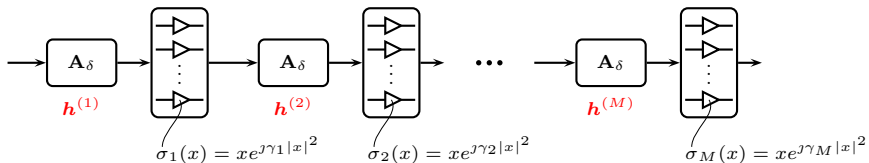
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Our approach

Joint optimization, pruning, and quantization of all chromatic-dispersion filters leads to **efficient digital backpropagation**, even with many steps.

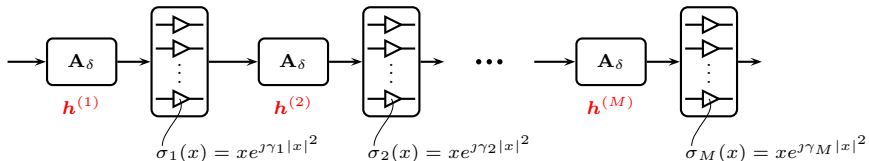
Learned Digital Backpropagation

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TensorFlow implementation of the computation graph $f_{\theta}(\mathbf{y})$:

Learned Digital Backpropagation

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Deep learning of parameters $\theta = \{\mathbf{h}^{(1)}, \dots, \mathbf{h}^{(M)}\}$:

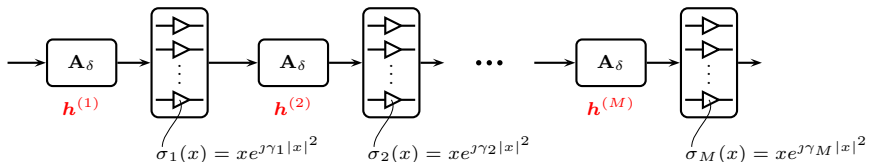
$$\min_{\theta} \sum_{i=1}^N \text{Loss}(f_{\theta}(\mathbf{y}^{(i)}), \mathbf{x}^{(i)}) \triangleq g(\theta)$$

mean squared error

using $\theta_{k+1} = \theta_k - \lambda \nabla_{\theta} g(\theta_k)$
 Adam optimizer, fixed learning rate

Learned Digital Backpropagation

TensorFlow implementation of the computation graph $f_{\theta}(\mathbf{y})$:



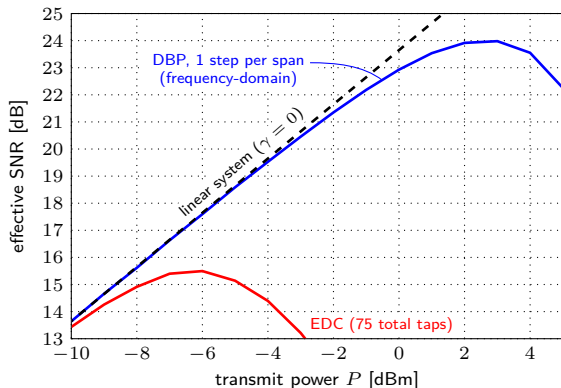
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Iteratively **prune (set to 0) outermost filter taps** during gradient descent

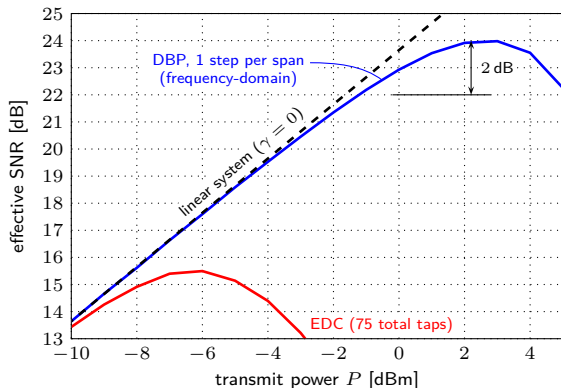
Revisiting Ip and Kahn (2008)



Parameters similar to [Ip and Kahn, 2008]:

- 25×80 km SSFM
- Gaussian modulation
- RRC pulses (0.1 roll-off)
- 10.7 Gbaud
- 2 samples/symbol processing
- single channel, single pol.

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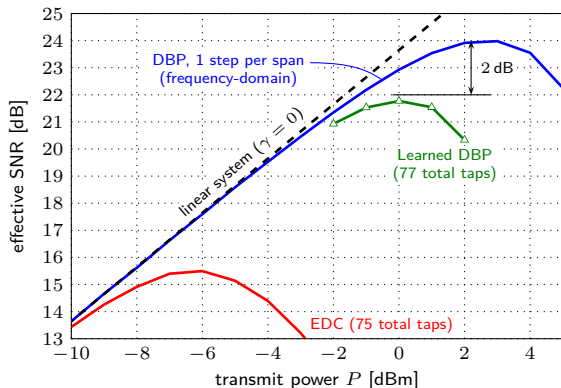


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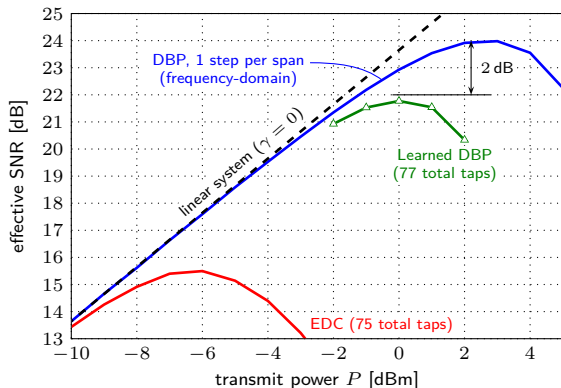


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- Learned approach uses **only 77 total taps**: alternate 5 and 3 taps/step and use **different** filter coefficients in all steps [Häger and Pfister, 2018a]

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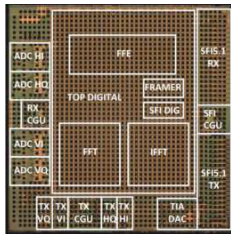


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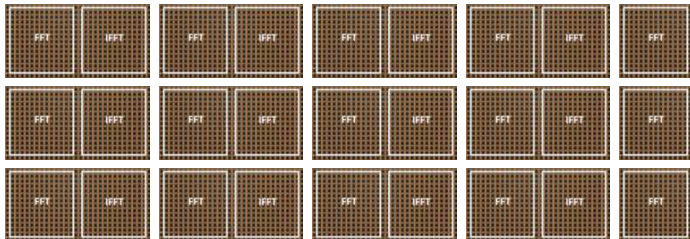
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- RRC pulses (0.1 roll-off)
- 10.7 Gbaud
- 2 samples/symbol processing
- single channel, single pol.

- \gg 1000 total taps (70 taps/step) \implies $> 100\times$ complexity of EDC
- Learned approach uses **only 77 total taps**: alternate 5 and 3 taps/step and use **different** filter coefficients in all steps [Häger and Pfister, 2018a]
- Can **outperform** "ideal DBP" in the nonlinear regime [Häger and Pfister, 2018b]

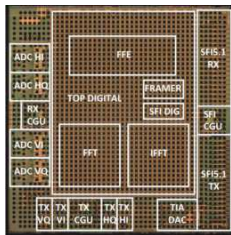
Real-Time ASIC Implementation



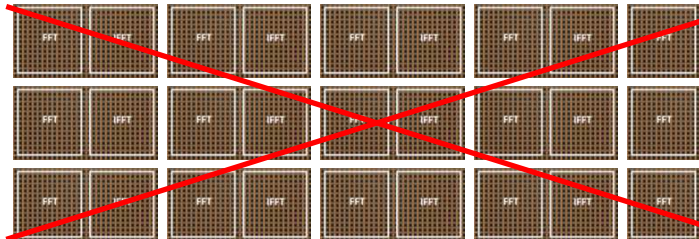
[Crivelli et al., 2014]



Real-Time ASIC Implementation



[Crivelli et al., 2014]

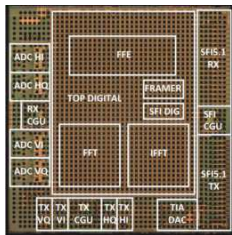


[Fougstedt et al., 2017], Time-domain digital back propagation: Algorithm and finite-precision implementation aspects, (*OFC*)

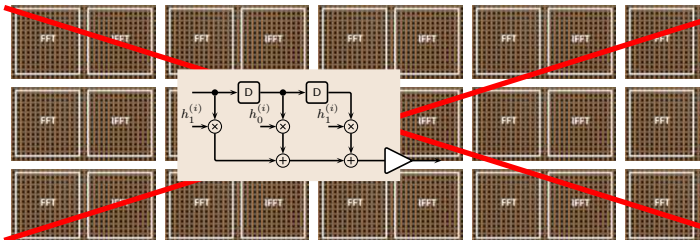
[Fougstedt et al., 2018], ASIC implementation of time-domain digital back propagation for coherent receivers, (*PTL*)

[Sherborne et al., 2018], On the impact of fixed point hardware for optical fiber nonlinearity compensation algorithms, (*JLT*)

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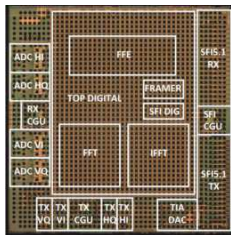


[Crivelli et al., 2014]

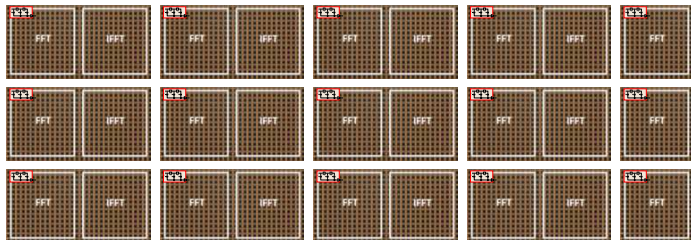


- Our linear steps are **very short symmetric FIR filters** (as few as **3 taps**)

Real-Time ASIC Implementation



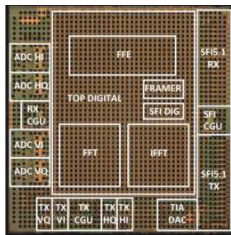
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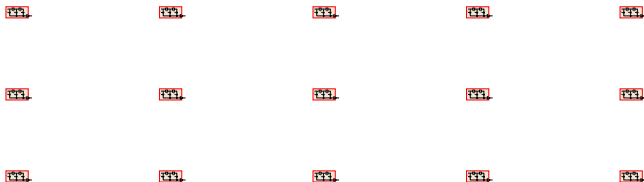
- Our linear steps are **very short symmetric FIR filters** (as few as **3 taps**)
- 28-nm ASIC at 416.7 MHz clock speed (40 GHz signal)
 - **Only 5-6 bit** filter coefficients via **learned quantization**
 - Hardware-friendly nonlinear steps (Taylor expansion)
 - All FIR filters are **fully reconfigurable**

[Fougstedt et al., 2018], ASIC implementation of time-domain digital backpropagation with deep-learned chromatic dispersion filters, (ECOC)

Real-Time ASIC Implementation



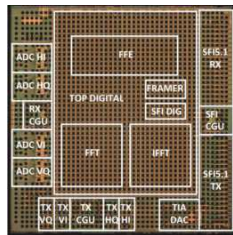
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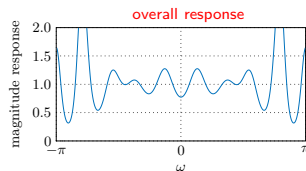
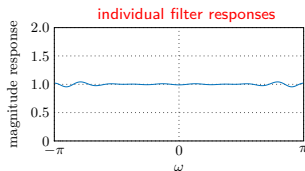
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 - All FIR filters are **fully reconfigurable**
- **< 2× power compared to EDC** [Crivelli et al., 2014, Pillai et al., 2014]

[Fougstedt et al., 2018], ASIC implementation of time-domain digital backpropagation with deep-learned chromatic dispersion filters, (ECOC)

Why Does The Learning Approach Work?

Previous work: design a single filter or filter pair and use it repeatedly.

⇒ Good overall response only possible with very long filters.



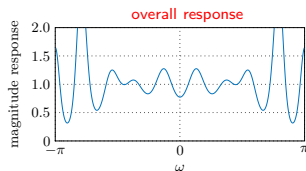
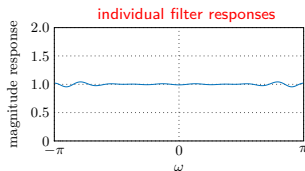
From [Ip and Kahn, 2009]:

- “We also note that [. . .] 70 taps, is much larger than expected”
- “This is due to amplitude ringing in the frequency domain”
- “Since backpropagation requires multiple iterations of the linear filter, amplitude distortion due to ringing accumulates (Goldfarb & Li, 2009)”

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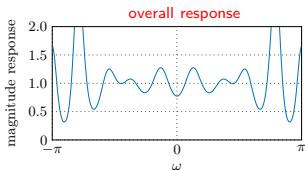
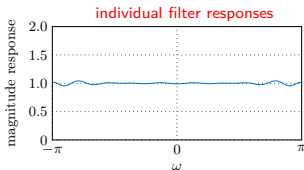
The learning approach uncovered that there is no such requirement!

[Lian, Häger, Pfister, 2018], What can machine learning teach us about communications? (*ITW*)

Why Does The Learning Approach Work?

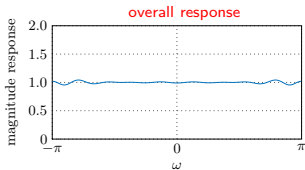
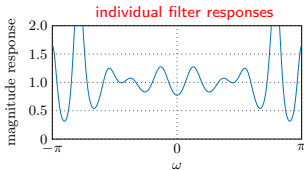
Previous work: design a single filter or filter pair and use it repeatedly.

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Sacrifice individual filter accuracy, but different response per step.

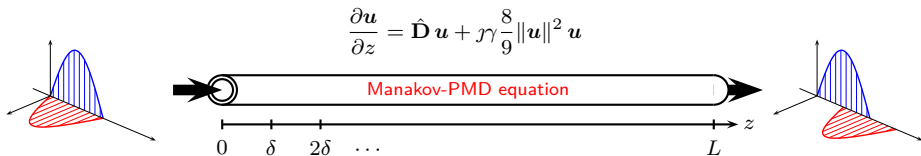
⇒ Good overall response even with very short filters by joint optimization.



Outline

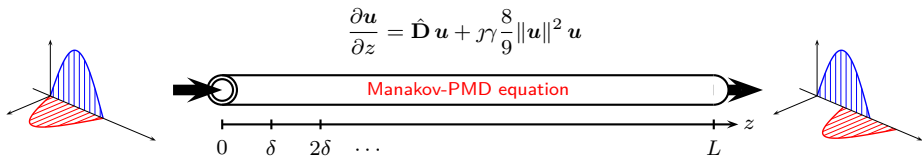
1. Machine Learning and Neural Networks for Communications
2. Model-Based Machine Learning for Fiber-Optic Communications
3. Learned Digital Backpropagation
4. Polarization-Dependent Effects
5. Conclusions

Evolution of Polarization-Multiplexed Signals

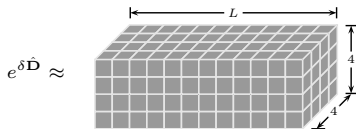
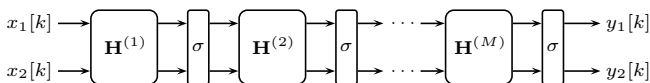


- Jones vector $\mathbf{u} \triangleq (u_1(t, z), u_2(t, z))^T$ with complex baseband signals
- linear operator $\hat{\mathbf{D}}$: attenuation, chromatic & polarization mode dispersion

Evolution of Polarization-Multiplexed Signals

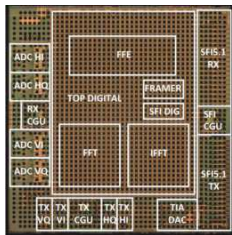


- Jones vector $\mathbf{u} \triangleq (u_1(t, z), u_2(t, z))^T$ with complex baseband signals
- **linear operator** $\hat{\mathbf{D}}$: attenuation, chromatic & polarization mode dispersion
- Split-step method: **alternate linear and nonlinear steps** $\sigma(x) = x e^{j\gamma \frac{8}{9} \delta \|x\|^2}$

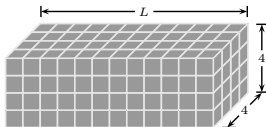


2 × 2 complex or
4 × 4 real MIMO filters \Rightarrow **complexity!**

Real-Time Compensation of Polarization Impairments

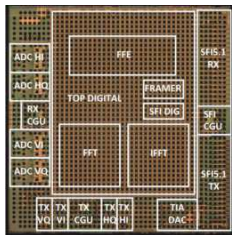


[Crivelli et al., 2014]

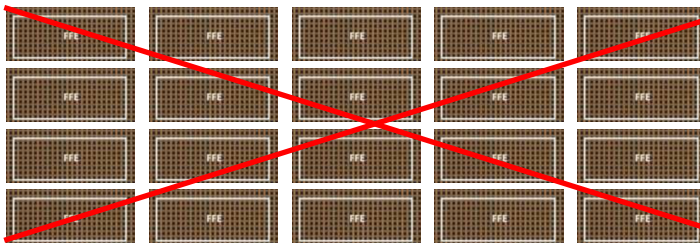


- **time-varying** effects (e.g., drifts) & a priori **unknown realizations**
- \implies **adaptive filtering** (via stochastic gradient descent) required

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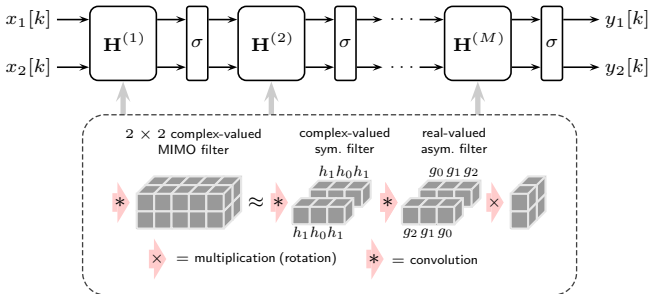
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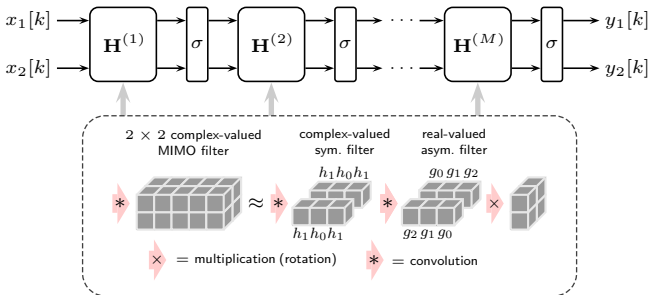
Using (and updating) **full MIMO filters** in each step is **not feasible**.

Our approach: Factorize each MIMO Filter



- 5-tap real-valued filters to approximate **first-order PMD (DGD)**
- **Memoryless rotations** $\begin{pmatrix} a & -b^* \\ b & a^* \end{pmatrix}$, where $a, b \in \mathbb{C}$ (4 real parameters)

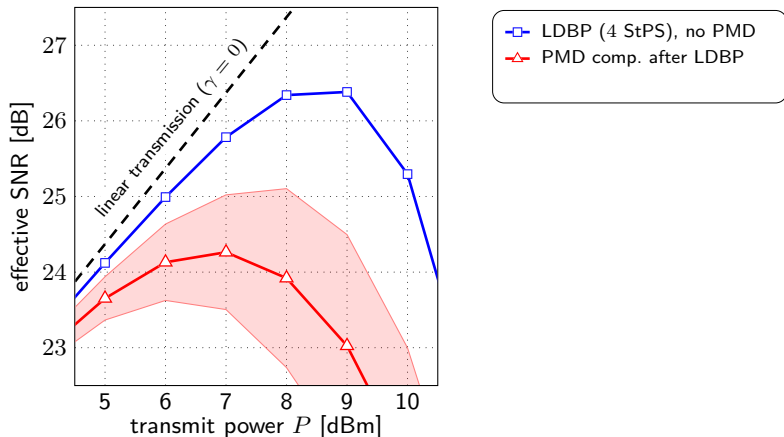
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- **Memoryless rotations** $\begin{pmatrix} a & -b^* \\ b & a^* \end{pmatrix}$, where $a, b \in \mathbb{C}$ (4 real parameters)
- **Assumes no knowledge** about PMD realizations or **accumulated PMD**
- **FIR-filter based!** **Avoids frequency-domain** (FFT-based) filtering

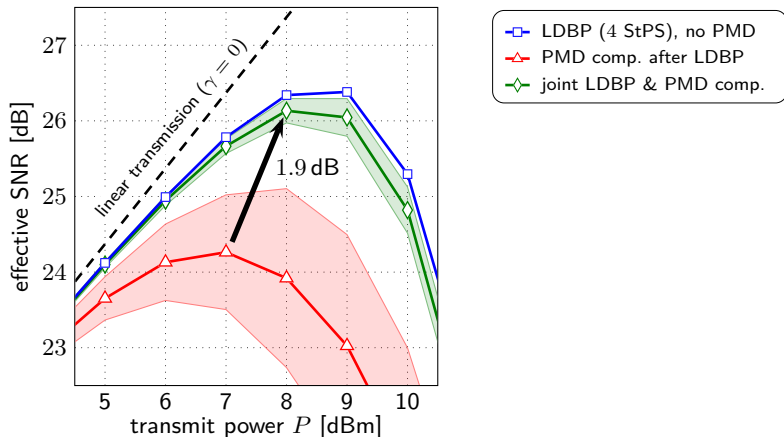
[Goroshko et al., 2016], Overcoming performance limitations of digital back propagation due to polarization mode dispersion, (*CTON*)
 [Czegledi et al., 2017], Digital backpropagation accounting for polarization-mode dispersion, (*Opt. Express*)
 [Liga et al., 2018], A PMD-adaptive DBP receiver based on SNR optimization, (*OFC*)

Results (32 Gbaud, 10×100 km, 0.2 ps/ $\sqrt{\text{km}}$ PMD)



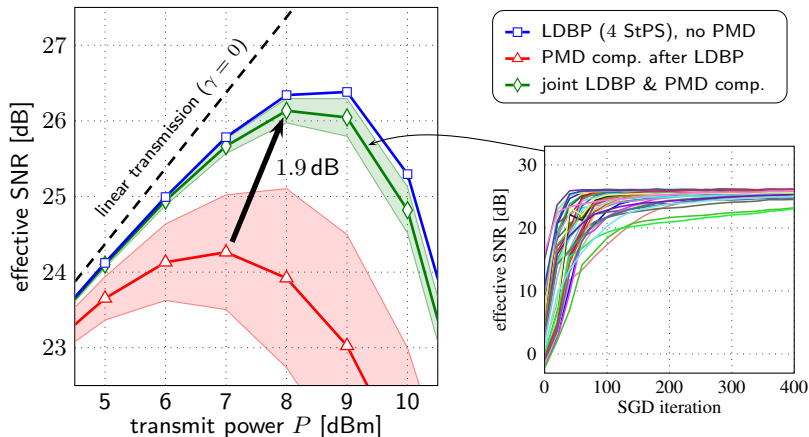
- Similar parameters & simulation setup compared to [Czegledi et al., 2016], results averaged over 40 PMD realizations

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- Similar parameters & simulation setup compared to [Czegledi et al., 2016], results averaged over 40 PMD realizations
- **Reliable convergence** “from scratch” + only 9 real parameters per step

Related and Recent Works

Learned digital backpropagation

- [Häger & Pfister, 2018], Nonlinear Interference Mitigation via Deep Neural Networks, (*OFC*)
- [Häger & Pfister, 2018], Deep Learning of the Nonlinear Schrödinger Equation in Fiber-Optic Communications, (*ISIT*)
- [Lian, Häger, Pfister, 2018], What can machine learning teach us about communications? (*ITW*)
- [Häger et al., 2019], Revisiting multi-step nonlinearity compensation with machine learning (*ECOC*)

Wideband & WDM signals (alternating filter banks and nonlinearities)

- [Häger and Pfister, 2018], Wideband time-domain digital backpropagation via subband processing and deep learning, (*ECOC*)

ASIC implementation & finite-precision aspects

- [Fougstedt et al., 2018], ASIC implementation of time-domain digital backpropagation with deep-learned chromatic dispersion filters, (*ECOC*)

Experimental demonstrations & implementation aspects (e.g., phase noise)

- [Sillekens et al., 2020], Experimental demonstration of learned time-domain digital back-propagation, (*arXiv*)
- [Bitachon et al., 2020], Deep learning based digital back propagation with polarization state rotation & phase noise invariance, (*OFC*)

The Bigger Picture

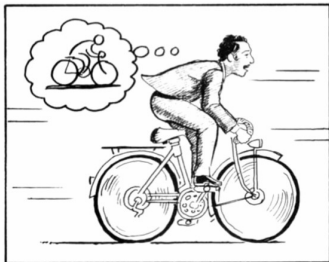
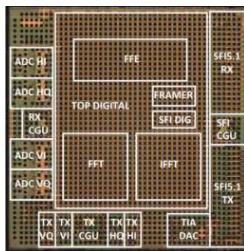


Figure 1. A World Model, from Scott McCloud's *Understanding Comics*. (McCloud, 1993; E, 2012)



[Crivelli et al., 2014]

- Optical receivers build models of their "environment"

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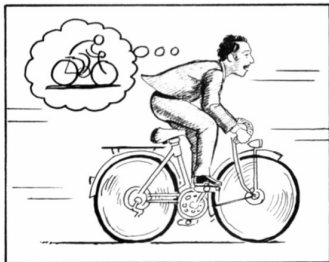
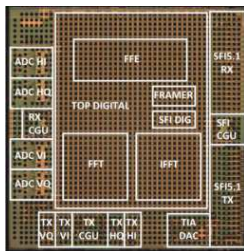


Figure 1. A World Model, from Scott McCloud's *Understanding Comics*. (McCloud, 1993; E, 2012)



[Crivelli et al., 2014]

- Optical receivers build models of their "environment"
- Currently these models are **linear** and/or **rigid** (non-adaptive)
- Interpretable **physics-based "multi-layer" models** for machine learning can be obtained by exploiting our existing domain knowledge

Conclusions

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neural-network-based ML

universal function approximators

good designs require
experience and fine-tuning

black boxes,
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Thank you!



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