

Density Evolution for Deterministic Generalized Product Codes with Higher-Order Modulation

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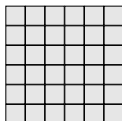
In This Talk ...

- **Deterministic** code construction that recovers product codes, staircase codes, and block-wise braided codes as special cases
- Rigorous **density evolution** analysis over **parallel** BECs
- **Application**: Bit mapper (interleaver) optimization for coded modulation

Introduction: Product Codes and Staircase Codes

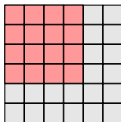
Introduction: Product Codes and Staircase Codes

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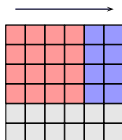
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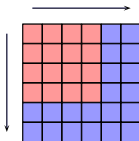
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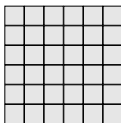
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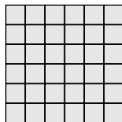
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each row/column is a codeword in
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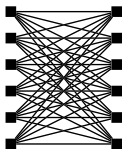
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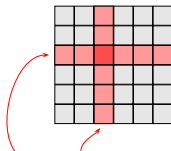
Tanner
graph



constraint node (CN) degree = component code length

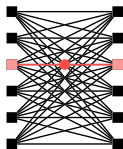
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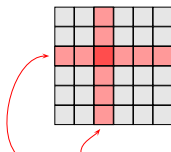


edge = degree-2 variable node (VN)

constraint node (CN) degree = component code length

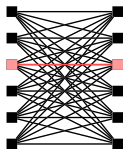
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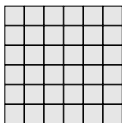
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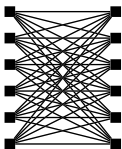
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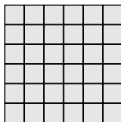


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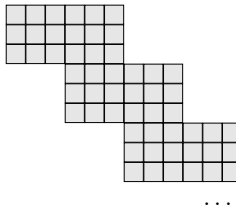


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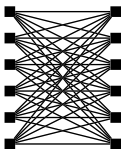
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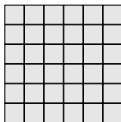


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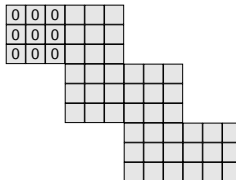


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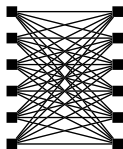


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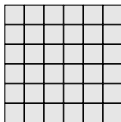
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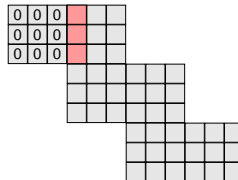


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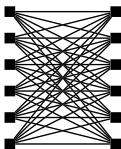
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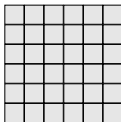


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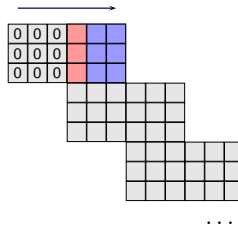


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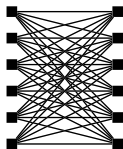
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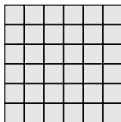


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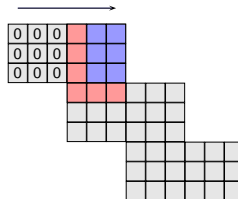


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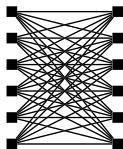


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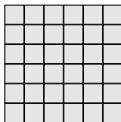
...

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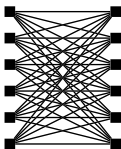


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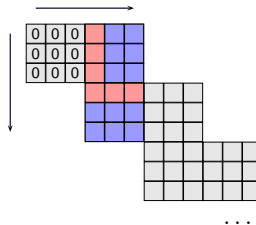
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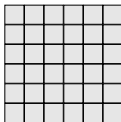


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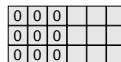


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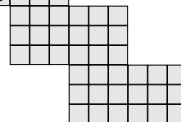
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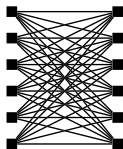


generalized product code (GPC)



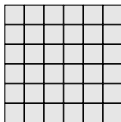
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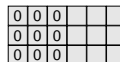


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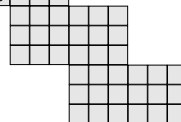
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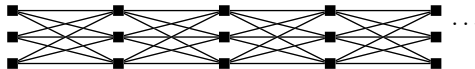
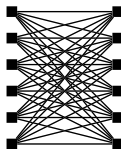


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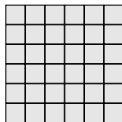
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Tanner graph

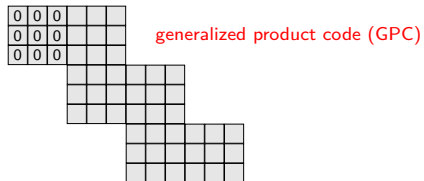


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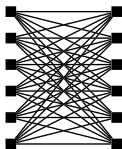
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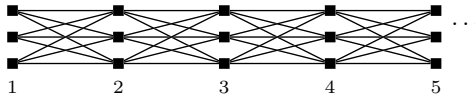


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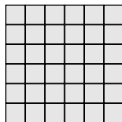
spatially-coupled code

positions: 1 2 3 4 5

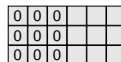


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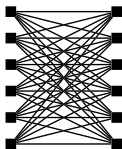


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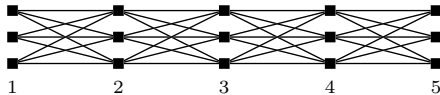


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Tanner graph

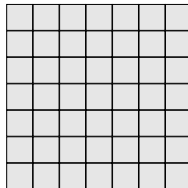
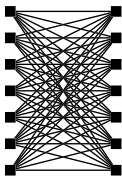


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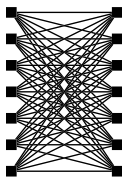


- **Deterministic** codes with fixed and structured Tanner graph

Iterative Bounded-Distance Decoding

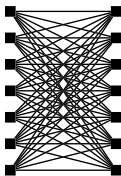


Iterative Bounded-Distance Decoding



| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | 1 | 0 | 1 | 0 | 1 | 0 |
| 0 | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 0 | 1 | 0 | 1 | 0 |
| 1 | 1 | 1 | 0 | 1 | 1 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | 0 | 0 | 1 | 1 | 1 |
| 0 | 1 | 0 | 0 | 0 | 1 | 1 |

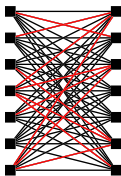
Iterative Bounded-Distance Decoding



| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | ? | 0 | ? | 0 | 1 | ? |
| ? | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 0 | ? | 0 | ? | ? |
| 1 | 1 | 1 | 0 | 1 | 1 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | ? | ? | 1 | 1 | ? |
| 0 | 1 | 0 | ? | 0 | 1 | 1 |

- Codeword transmission over the **BEC** with erasure probability p

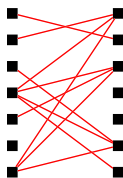
Iterative Bounded-Distance Decoding



| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | ? | 0 | ? | 0 | 1 | ? |
| ? | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 0 | ? | 0 | ? | ? |
| 1 | 1 | 1 | 0 | 1 | 1 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 1 |
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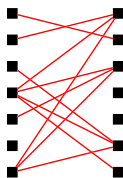


residual graph

| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | ? | 0 | ? | 0 | 1 | ? |
| ? | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 0 | ? | 0 | ? | ? |
| 1 | 1 | 1 | 0 | 1 | 1 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | ? | ? | 1 | 1 | ? |
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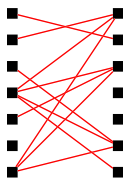


residual graph

| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | ? | 0 | ? | 0 | 1 | ? |
| ? | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 0 | ? | 0 | ? | ? |
| 1 | 1 | 1 | 0 | 1 | 1 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | ? | ? | 1 | 1 | ? |
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- Codeword transmission over the **BEC** with erasure probability p
- Each CN corresponds to **t -erasure correcting component code**

Iterative Bounded-Distance Decoding



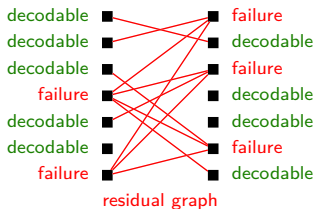
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| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | ? | 0 | ? | 0 | 1 | ? |
| ? | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 0 | ? | 0 | ? | ? |
| 1 | 1 | 1 | 0 | 1 | 1 | 0 |
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Iterative Bounded-Distance Decoding

1st iteration ($t = 2$)

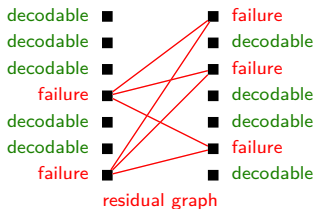


| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | ? | 0 | ? | 0 | 1 | ? |
| ? | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 0 | ? | 0 | ? | ? |
| 1 | 1 | 1 | 0 | 1 | 1 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | ? | ? | 1 | 1 | ? |
| 0 | 1 | 0 | ? | 0 | 1 | 1 |

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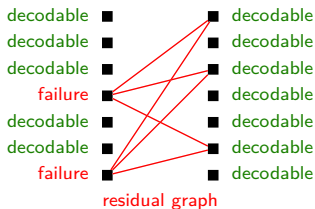


| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | 1 | 0 | ? | 0 | 1 | ? |
| 0 | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 0 | ? | 0 | 1 | ? |
| 1 | 1 | 1 | 0 | 1 | 1 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | 0 | ? | 1 | 1 | ? |
| 0 | 1 | 0 | 0 | 0 | 1 | 1 |

- Codeword transmission over the **BEC** with erasure probability p
- Each CN corresponds to **t -erasure correcting component code**
- ℓ iterations of **bounded-distance decoding** = **peeling** of vertices with degree $\leq t$ (in parallel)

Iterative Bounded-Distance Decoding

2nd iteration ($t = 2$)



| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | 1 | 0 | ? | 0 | 1 | ? |
| 0 | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 0 | ? | 0 | 1 | ? |
| 1 | 1 | 1 | 0 | 1 | 1 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | 0 | ? | 1 | 1 | ? |
| 0 | 1 | 0 | 0 | 0 | 1 | 1 |

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| | |
|-------------|-------------|
| decodable ■ | ■ decodable |
| decodable ■ | ■ decodable |
| decodable ■ | ■ decodable |
| failure ■ | ■ decodable |
| decodable ■ | ■ decodable |
| decodable ■ | ■ decodable |
| failure ■ | ■ decodable |

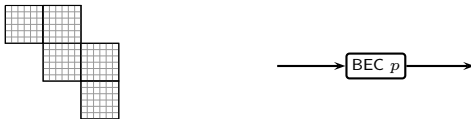
residual graph

| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | 1 | 0 | 1 | 0 | 1 | 0 |
| 0 | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 0 | 1 | 0 | 1 | 0 |
| 1 | 1 | 1 | 0 | 1 | 1 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 1 |
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| 0 | 1 | 0 | 0 | 0 | 1 | 1 |

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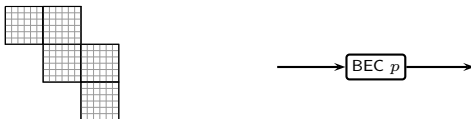
Asymptotic Performance Prediction

Asymptotic Performance Prediction



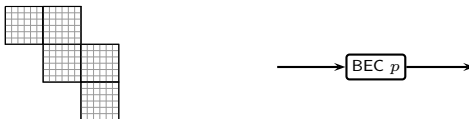
- Example: **staircase code** with a fixed component code

Asymptotic Performance Prediction

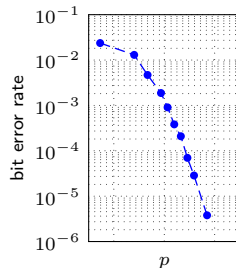


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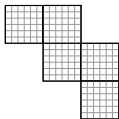
Asymptotic Performance Prediction



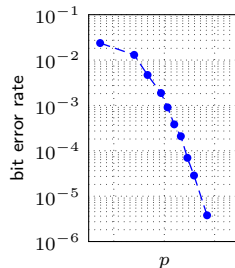
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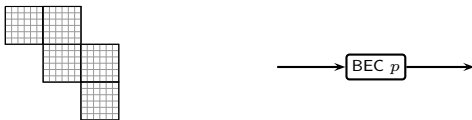
Asymptotic Performance Prediction



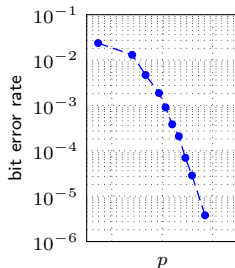
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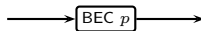
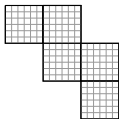
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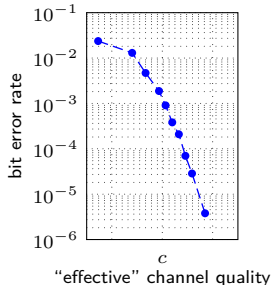
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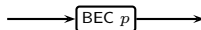
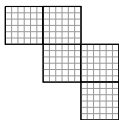
Asymptotic Performance Prediction



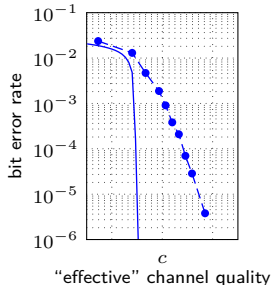
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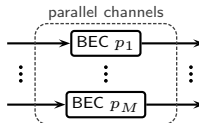
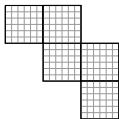
Asymptotic Performance Prediction



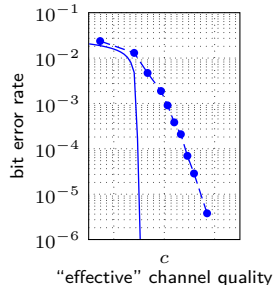
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Asymptotic Performance Prediction



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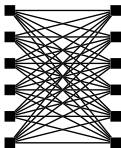
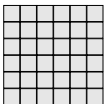


Main contribution

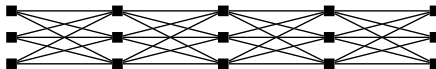
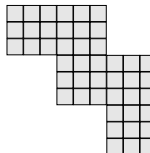
Generalization to **parallel** BECs with different erasure probabilities

Parametrized Construction of Generalized Product Codes

product codes



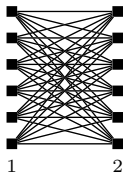
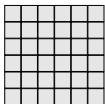
staircase codes



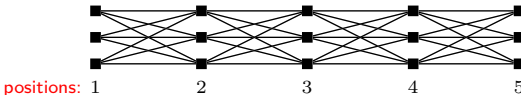
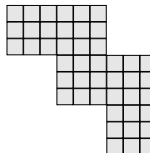
positions: 1 2 3 4 5

Parametrized Construction of Generalized Product Codes

product codes

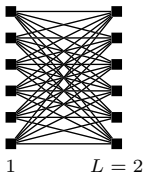
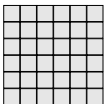


staircase codes

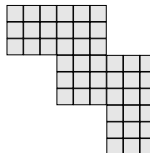


Parametrized Construction of Generalized Product Codes

product codes

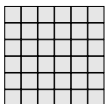


staircase codes

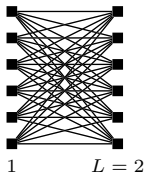
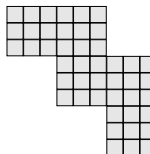


Parametrized Construction of Generalized Product Codes

product codes



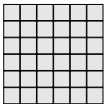
staircase codes



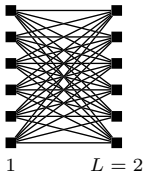
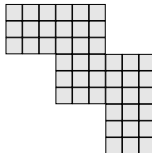
η : symmetric $L \times L$ matrix that defines **graph connectivity**

Parametrized Construction of Generalized Product Codes

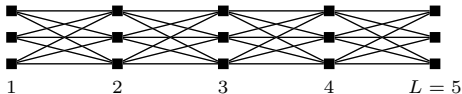
product codes



staircase codes



positions:

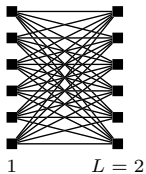
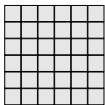


$$\eta = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$$

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Parametrized Construction of Generalized Product Codes

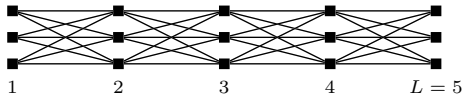
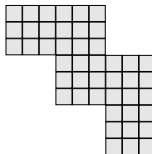
product codes



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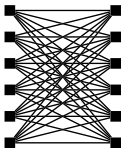
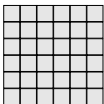
staircase codes



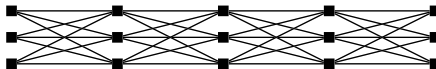
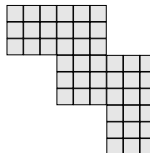
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Parametrized Construction of Generalized Product Codes

product codes

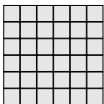


staircase codes

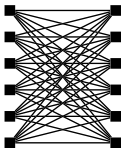
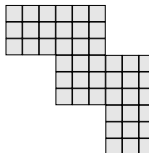


Parametrized Construction of Generalized Product Codes

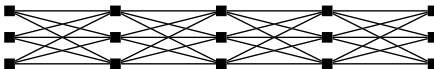
product codes



staircase codes

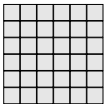


n : "problem size", proportional to
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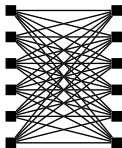
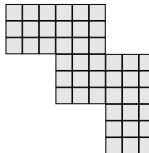


Parametrized Construction of Generalized Product Codes

product codes



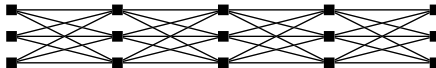
staircase codes



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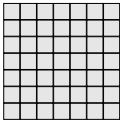


increasing n

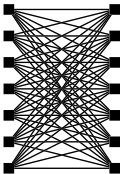
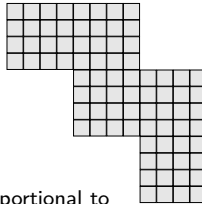


Parametrized Construction of Generalized Product Codes

product codes



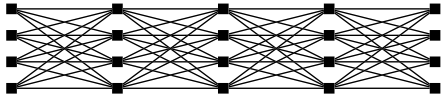
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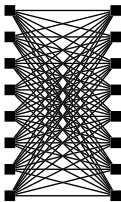
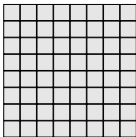


increasing n

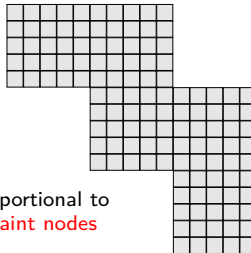


Parametrized Construction of Generalized Product Codes

product codes



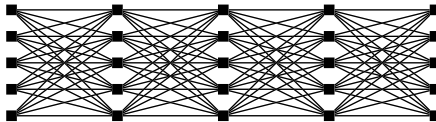
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Component Code Mixtures

Component Code Mixtures

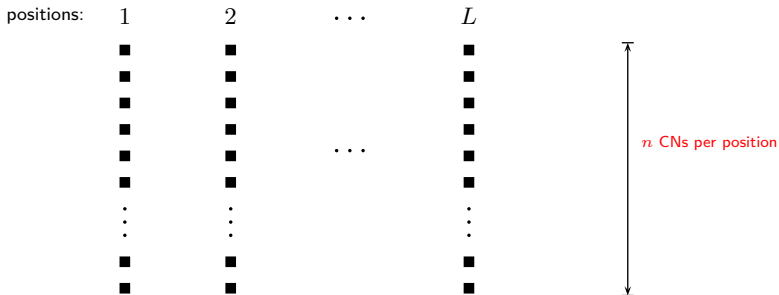
- For each position $i \in \{1, 2, \dots, L\}$, let $\boldsymbol{\tau}(i) = (\tau_1(i), \tau_2(i), \dots, \tau_{t_{\max}}(i))^T$ be a **probability vector/distribution** ($\sum_t \tau_t(i) = 1$ and $\tau_t(i) \geq 0$ for all i)

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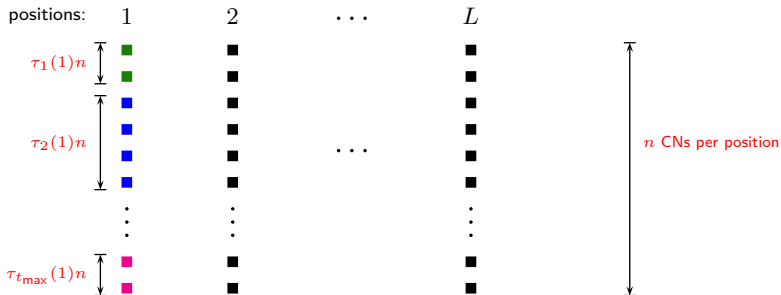
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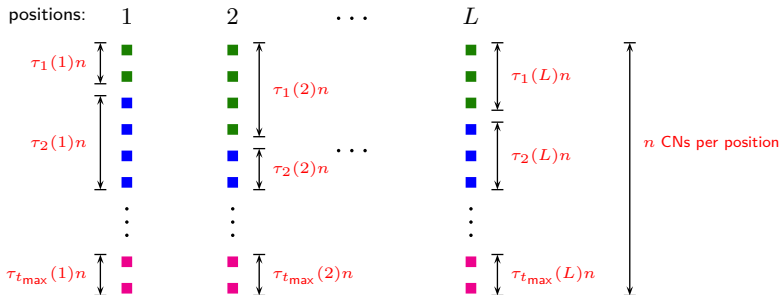
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Component Code Mixtures

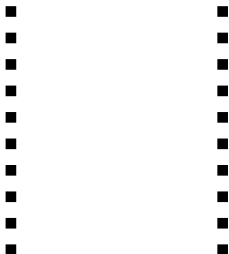
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- Example: **irregular product codes**, where $\boldsymbol{\eta} = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$ (i.e., $L = 2$) and
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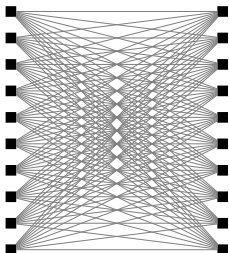
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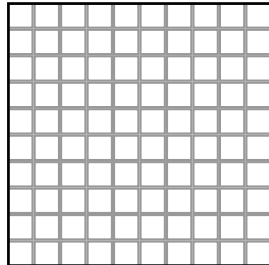
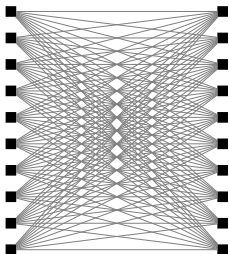
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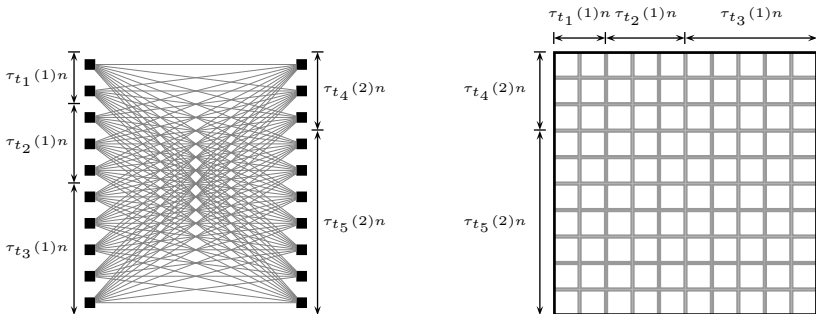
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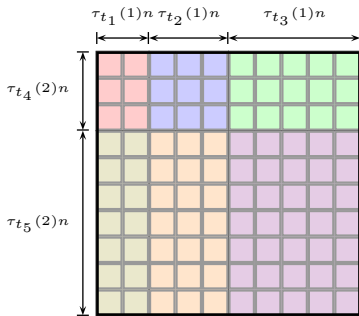
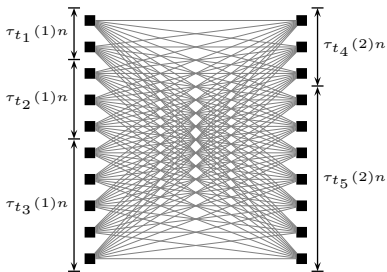
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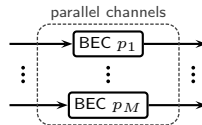
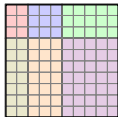
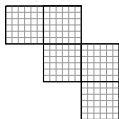
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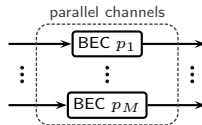
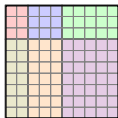
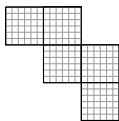


$K = 6$ distinct VN classes

Bit Mapper

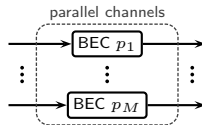
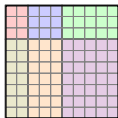
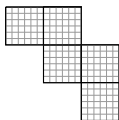


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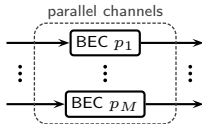
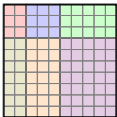
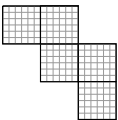
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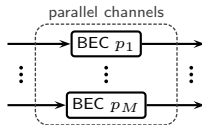
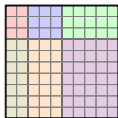
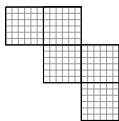
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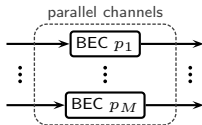
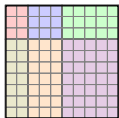
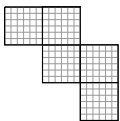
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- Example: **baseline/uniform bit mapper**, where $a_{k,q} = 1/M$ for all k, q
 \implies all virtual BECs are the same

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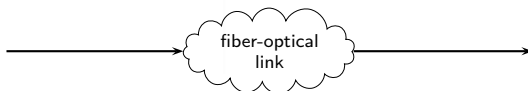
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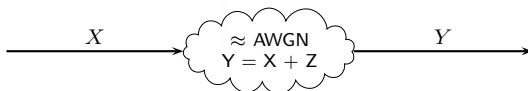
$$\bar{c} = \sup\{c > 0 \mid \lim_{\ell \rightarrow \infty} x_{i,t}^{(\ell)} = 0 \text{ for all } i, t\}$$

Spectrally-Efficient Communication

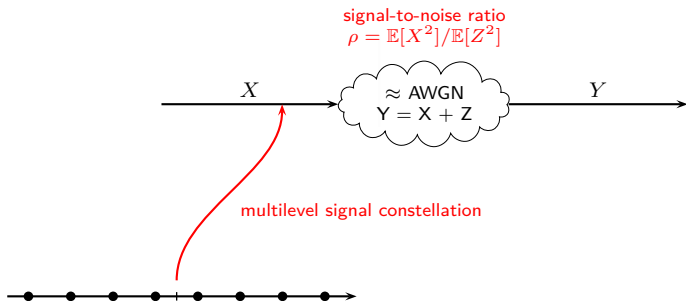


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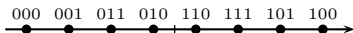
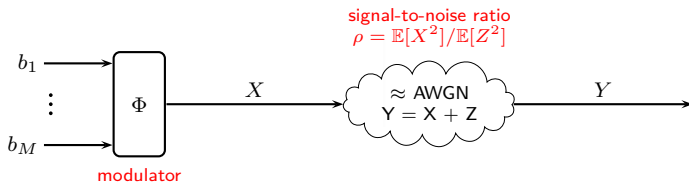
signal-to-noise ratio
 $\rho = \mathbb{E}[X^2]/\mathbb{E}[Z^2]$



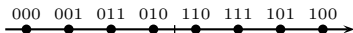
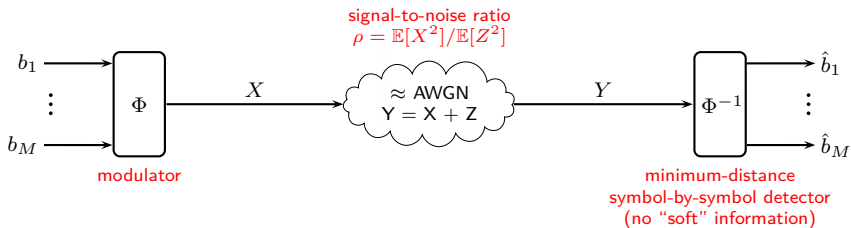
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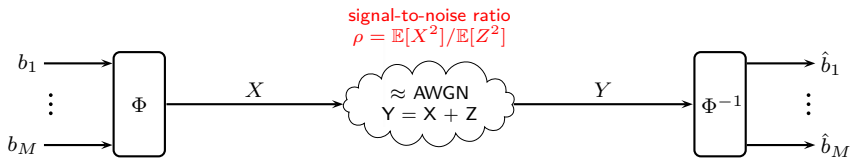
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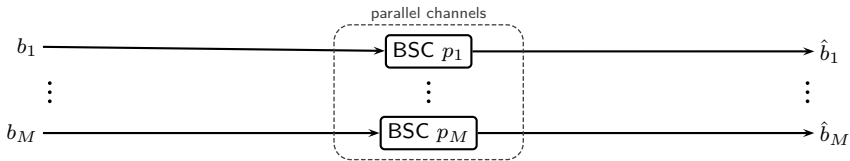
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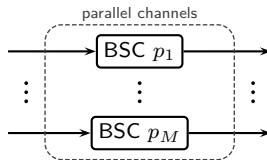


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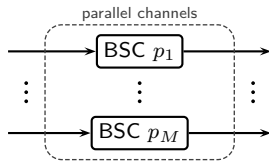
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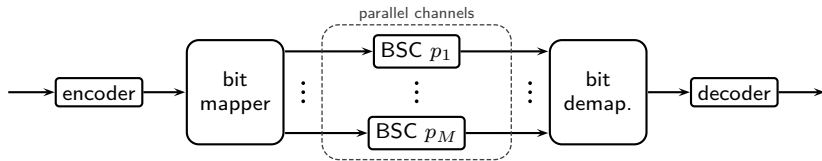
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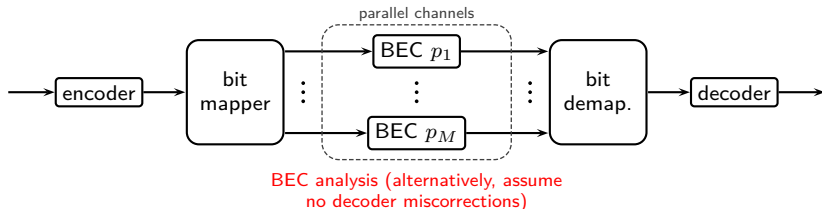
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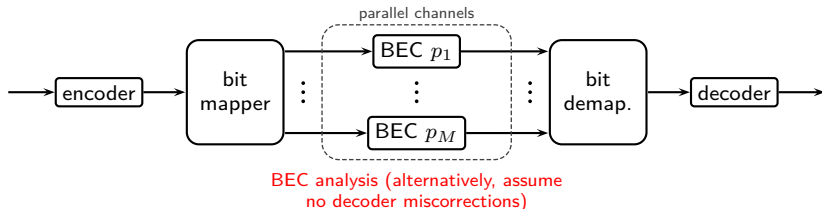
Spectrally-Efficient Communication



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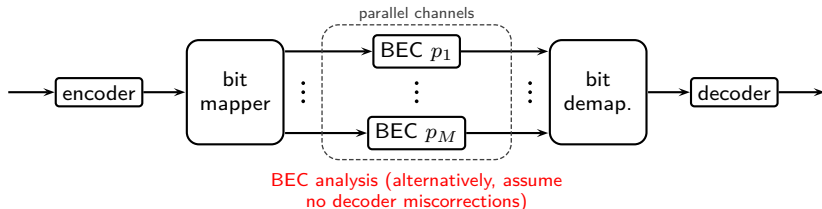


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- Example: threshold $\bar{c} = 10.63$ and $M = 4$. For $n = 1600$, **waterfall region expected** at $\rho = \bar{p}^{-1}(c/n) \approx 26.11$ dB

Bit Mapper Optimization

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Problem Formulation ([Richter et al., 2007], [Cheng et al., 2012], ...)

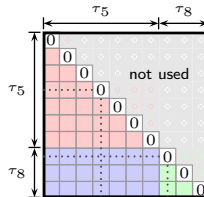
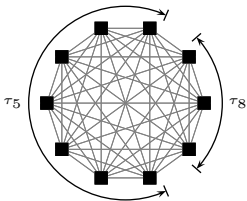
Optimize the bit mapper A for a given code and signal constellation

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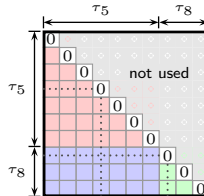
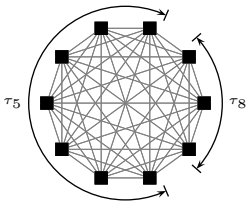


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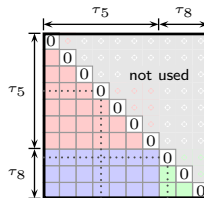
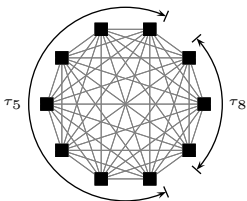
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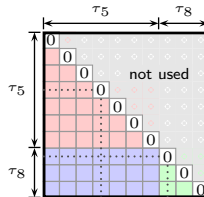
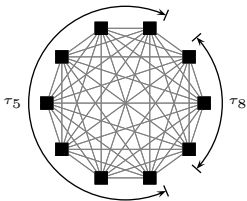
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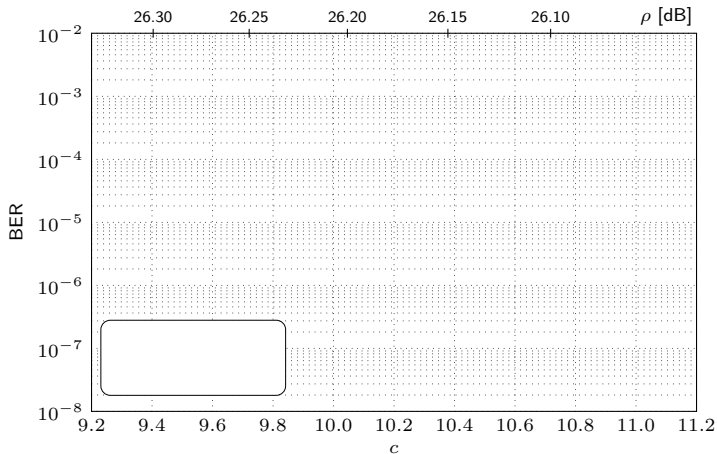
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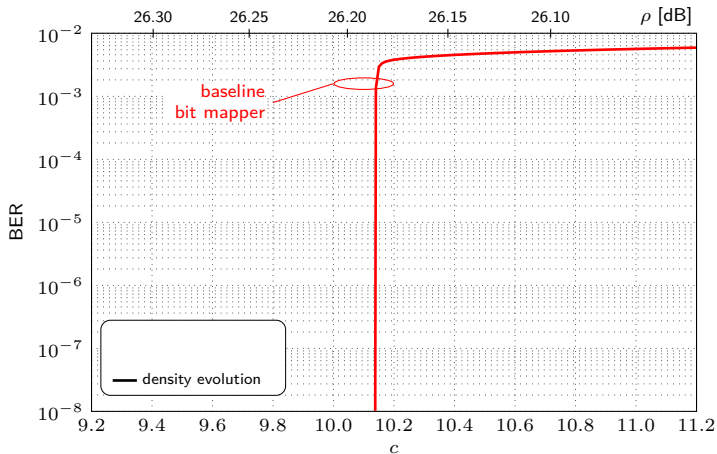


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Results

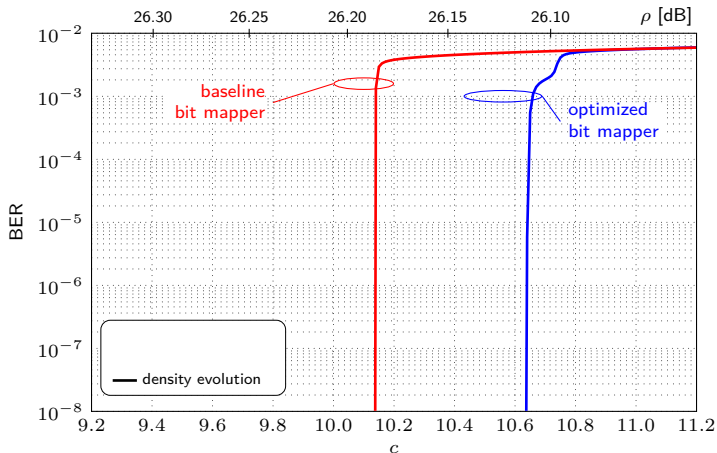


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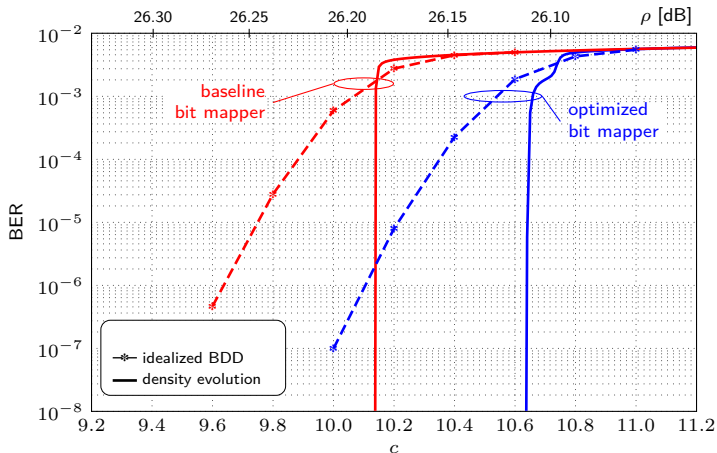


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most reliable least reliable

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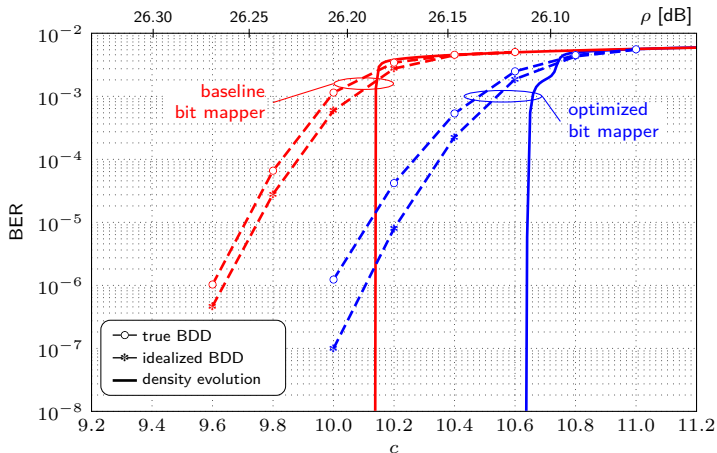


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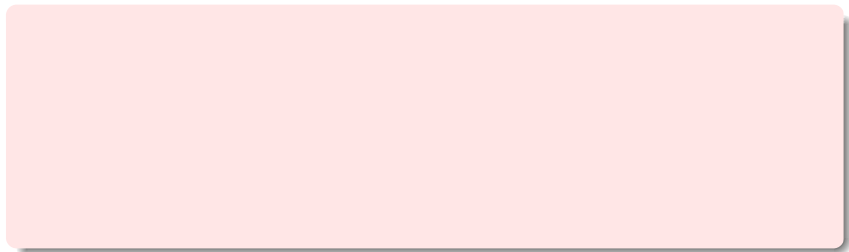


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Thank you!



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